

8 + 3

$$z^{-1} = \frac{-1}{z} \quad \frac{-15}{z} = \frac{5}{z} b$$

National Tsing Hua University  
Department of Electrical Engineering  
EE3660 Introduction to Digital Signal Processing, Spring 2020

Midterm Exam (100%)  
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$$-2 + 11z^{-1} = a(1 + 2z^{-1}) + b(z^{-1})$$

$z = 2$

Note: Detailed derivations are required to obtain a full score for each problem.

1. (15%) Consider a system which is described by

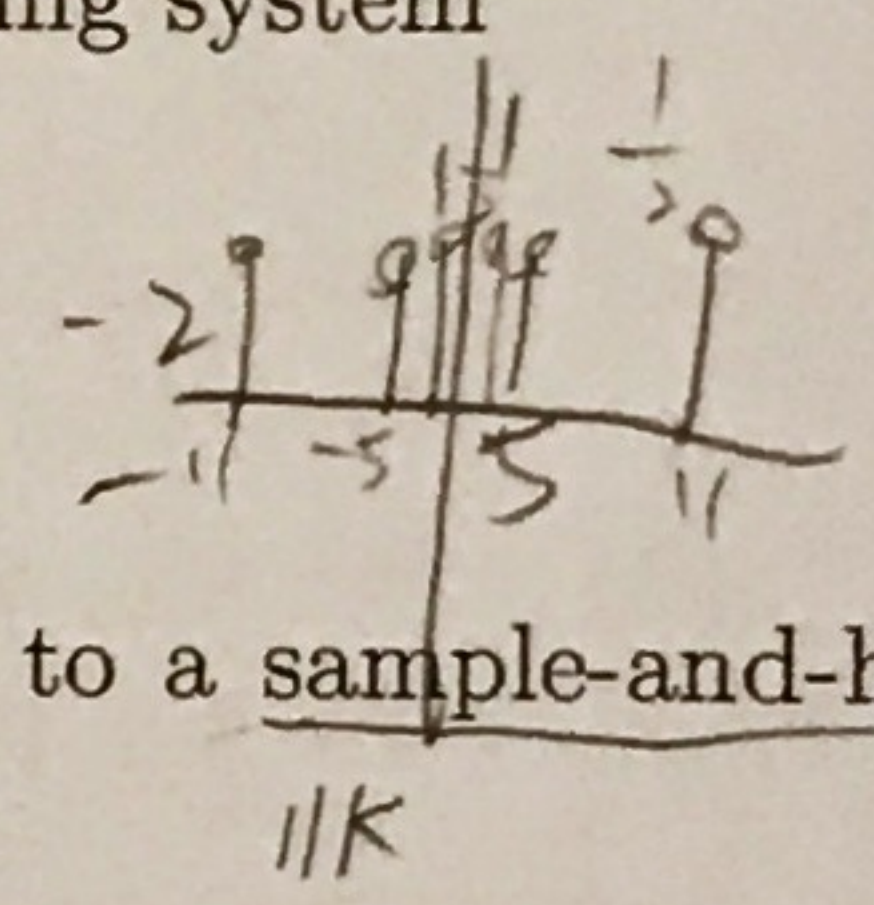
$$2y[n] + 3y[n-1] - 2y[n-2] = -2x[n] + 11x[n-1]$$

$$\begin{matrix} 2 & -1 \\ 1 & 2 \end{matrix} \quad \begin{matrix} (-z^{-1}) \\ (1 + 2z^{-1}) \end{matrix}$$

- (a) (5%) Determine the region of convergence for this system to be causal and also find the corresponding impulse response  $h[n]$ .
- (b) (5%) Determine the region of convergence for this system to be stable and also find the corresponding impulse response  $h[n]$ .
- (c) (5%) Convert this system into a minimum-phase one and determine the system function  $H_{\min}(z)$ .

2. (15%) Consider a sinusoidal signal  $x_c(t) = \cos(2\pi F_0 t)$  with  $F_0 = 11$  kHz. It is sampled at different rates of  $F_s$  and then reconstructed as  $y_r(t)$  by the corresponding system

$$G(j\Omega) = \begin{cases} \frac{\sin(\Omega T/2)}{\Omega/2} e^{-j\Omega T/2}, & |\Omega| < \pi/T \\ 0, & \text{otherwise} \end{cases}$$



where  $T = 1/F_s$ . Note that this reconstruction system is equivalent to a sample-and-hold DAC followed by an ideal lowpass filter.

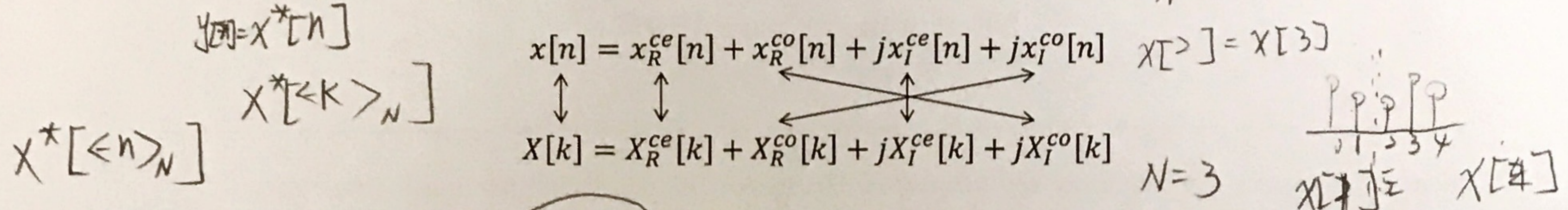
- (a) (5%) Determine  $y_r(t)$  if  $F_s = 33$  kHz.
- (b) (5%) Determine  $y_r(t)$  if  $F_s = 6$  kHz.
- (c) (5%) Consider a slightly complex scenario. Use  $x_c(t)$  as a carrier signal to modulate a low-frequency signal  $s(t) = \cos(2\pi F_1 t)$  where  $F_1 = 44$  Hz. And sample the modulated signal  $x_m(t) = s(t) \cdot x_c(t)$  at a rate of  $F_s = 44$  kHz. Then process the sampled signal  $x_m[n]$  by an all-pass digital filter  $H(e^{j\omega}) = e^{-j4\omega^3/\pi^2}$ . Finally, reconstruct  $y_r(t)$  from the filtered signal using the corresponding  $G(j\Omega)$ . Now approximate  $y_r(t)$  as accurate as you can. (Hint: use group delay to approximate the time delay of  $s(t)$ .)

3. (10%) The 8-point DFT of an 8-point sequence  $x[n]$  is given by

$$X[k] = \{1, 2 + j, 4 - j, 8 + j, 16, 8 - j, 4 - j, 2 - j\}$$

Determine the DFT of each of the following sequences.

- (a) (2%)  $x_1[n] = x[\langle n - 2 \rangle_8]$ .
- (b) (4%)  $x_2[l] = \sum_{n=0}^7 x[n]x^*[\langle n - l \rangle_8]$  (circular autocorrelation).
- (c) (4%) Real-valued  $x_3[n]$  and  $x_4[n]$  such that  $x[n] = x_3[n] + jx_4[n]$ . You may need the symmetry properties as below.



- 4. (15%) Let  $X[k]$  be a real-valued  $N$ -point DFT for an  $N$ -point sequence  $x[n]$ .
  - (a) (4%) Show that  $x[\frac{N-1}{2}] = x^*[\frac{N+1}{2}]$  if  $N$  is odd.
  - (b) (5%) Show that  $|x[n]| = |x[\langle -n \rangle_N]|$ .
  - (c) (6%) If  $X[k]$  satisfies the condition  $X[k] = X[\langle k + M \rangle_N]$  where  $N = lM$  and  $l$  is an integer, show that  $x[n] = 0$  if  $l \nmid n$ .
- 5. (15%) Complete the following example for computational Fourier analysis.
  - (a) (3%) For a right-sided exponential function  $x_c(t) = 2^{-t/T}u(t)$ , determine its CTFT  $X_c(j\Omega)$  given that  $e^{-at}u(t) \xrightarrow{\mathcal{F}} \frac{1}{a+j\Omega}$ .
  - (b) (4%) Sample  $x_c(t)$  with  $t = nT$  to form a discrete-time signal  $x[n] = x_c(nT)$  and then determine its DTFT  $X(e^{j\omega})$ . Also briefly state the relationship between  $X(e^{j\omega})$  and  $X_c(j\Omega)$ .
  - (c) (4%) Sample the frequency components of  $X(e^{j\omega})$  as four-point DFT  $X_1[k] = X(e^{j2\pi k/4})$  where  $k = 0, 1, \dots, 3$ . Determine its IDFT  $x_1[n]$  which is a finite-length signal now.
  - (d) (4%) Now we want to have a detailed evaluation of the DTFT of  $x_1[n]$ . Explain how to use DFT computation to accurately derive  $X_1(e^{j\omega})$  at  $\omega = 2\pi k/1024$  where  $k = 0, 1, \dots, 1023$ .
- 6. (10%) Consider a highpass linear-phase FIR filter design using the Kaiser window  $w[n] = \frac{I_0(\beta\sqrt{1-[(n-\alpha)/\alpha]^2})}{I_0(\beta)}$  where  $0 \leq n \leq M$  and  $I_0(x)$  is the zeroth-order modified Bessel function. The specifications are  $\omega_p = 0.6\pi$ ,  $\omega_s = 0.5\pi$ , and  $\delta_s = 0.01$ . Determine the following design parameters to minimize the filter length: (a) (3%)  $\beta$ , (b) (3%) cut-off frequency  $\omega_c$ , and (c) (4%) window (filter) length  $L$ . You may need the following empirical relations:

$$\beta = \begin{cases} 0, & A < 21 \\ 0.5842(A - 21)^{0.4} + 0.07886(A - 21), & 21 \leq A \leq 50 \\ 0.1102(A - 8.7), & A > 50 \end{cases},$$

$$M = \frac{A - 8}{2.285 \Delta \omega} \tag{1}$$

- 7. (20%) Consider a MATLAB FFT function **myfft512(x)** which performs exactly 512-point DFT of a 512-point input vector **x**. It can be used for efficient computation. For example,

the IDFT  $y$  of a given 512-point DFT  $Y$  (as a row vector) can be derived by the following MATLAB code.

```
Yflip= [Y(1) Y(512 : -1 : 2)];  
y= myfft512(Yflip)/512;
```

Write down efficient MATLAB codes for the following algorithms using only the `myfft512` function and simple arithmetic/indexing/padding operations. All 1D vectors are row-wise arranged.

- (a) (10%) Linear convolution. Given a 400-point  $x_1$  and a 500-point  $x_2$ , derive its linear convolution  $x_3$  (899-point) through multiplication in DFT domain. You may consider the overlap-save or overlap-add method.
- (b) (5%) Time-domain resolution scaling-up. Given a real-valued 63-point  $x_4$ , derive its  $8\times$  scaled-up real-valued signal  $x_5$  (504-point) by first applying 64-point DFT, then padding zeros, e.g. `zeros(1, k)`, in DFT domain, and finally performing 512-point IDFT.
- (c) (5%) Efficient DFT for real-valued sequences. Given two 512-point real-valued sequences  $x_6$  and  $x_7$ , derive their 512-point DFTs  $X_6$  and  $X_7$ , respectively, by calling `myfft512(...)` only once. You may need the `real` and `imag` functions to extract the real and imaginary parts. You may also need the symmetry properties in the problem 3.