

HW6 Program Assignment

TA: 陳永泰

P1 (10%)

Consider again the inverse DFT given in (8.2).

$$X[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-kn}, \quad n = 0, 1, 2, \dots, N-1 \quad (8.2)$$

(a) (5%) Replace k by $\langle -k \rangle_N$ in (8.2) and show that the resulting summation is a DFT expression, that

is, IDFT $\{X[k]\} = \frac{1}{N} \text{DFT} \{X[\langle -k \rangle_N]\}$

ans:

$$\begin{aligned} x[n] &= \frac{1}{N} \sum_{k=0}^{N-1} X[-k] W_N^{-n(-k)} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} X[-k] W_N^{-n(N-k)} = \frac{1}{N} \sum_{k=0}^{N-1} X[-k] W_N^{-nk} \\ &= \text{DFT}\{X[-k]\} \end{aligned}$$

(b) (5%) Develop a MATLAB function $\mathbf{x} = \text{IDFT}(\mathbf{X}, N)$ using the `fft` function that uses the above approach. Verify your function on signal $x[n] = \{1, 2, 3, 4, 5, 6, 7, 8\}$.

```
% see function IDFT() at the end of the script
close all; clc; clear all
xn = 1:8;
X = fft(xn);
xr = IDFT(X,length(xn))
```

xr = 1x8
1 2 3 4 5 6 7 8

P2 (15%)

In this problem we will investigate differences in the speeds of DFT and FFT algorithms when stored twiddle factors are used.

(a) Write a function **W = dft_matrix(N)** that computes the DFT matrix W_N given in (8.8).

```
% see function DFTmtx()
n = 8;
D = dft_matrix(n)

D = 8x8 complex
    1.0000 + 0.0000i   1.0000 + 0.0000i   1.0000 + 0.0000i   1.0000 + 0.0000i ...

```

```

1.0000 + 0.0000i  0.7071 - 0.7071i  0.0000 - 1.0000i  -0.7071 - 0.7071i
1.0000 + 0.0000i  0.0000 - 1.0000i  -1.0000 - 0.0000i  -0.0000 + 1.0000i
1.0000 + 0.0000i  -0.7071 - 0.7071i  -0.0000 + 1.0000i  0.7071 - 0.7071i
1.0000 + 0.0000i  -1.0000 - 0.0000i  1.0000 + 0.0000i  -1.0000 - 0.0000i
1.0000 + 0.0000i  -0.7071 + 0.7071i  0.0000 - 1.0000i  0.7071 + 0.7071i
1.0000 + 0.0000i  -0.0000 + 1.0000i  -1.0000 - 0.0000i  0.0000 - 1.0000i
1.0000 + 0.0000i  0.7071 + 0.7071i  -0.0000 + 1.0000i  -0.7071 + 0.7071i

```

(b) (5%) Write a function **X = dftdirect_m(x,W)** that modifies the **dftdirect** function using the matrix W from (a). Using the **tic** and **toc** functions compare computation times for the **dftdirect** and **dftdirect_m** function for N = 128, 256, 512, and 1024. For this purpose generate an N-point complex-valued signal as **x = randn(1,N) + 1j*randn(1,N)**.

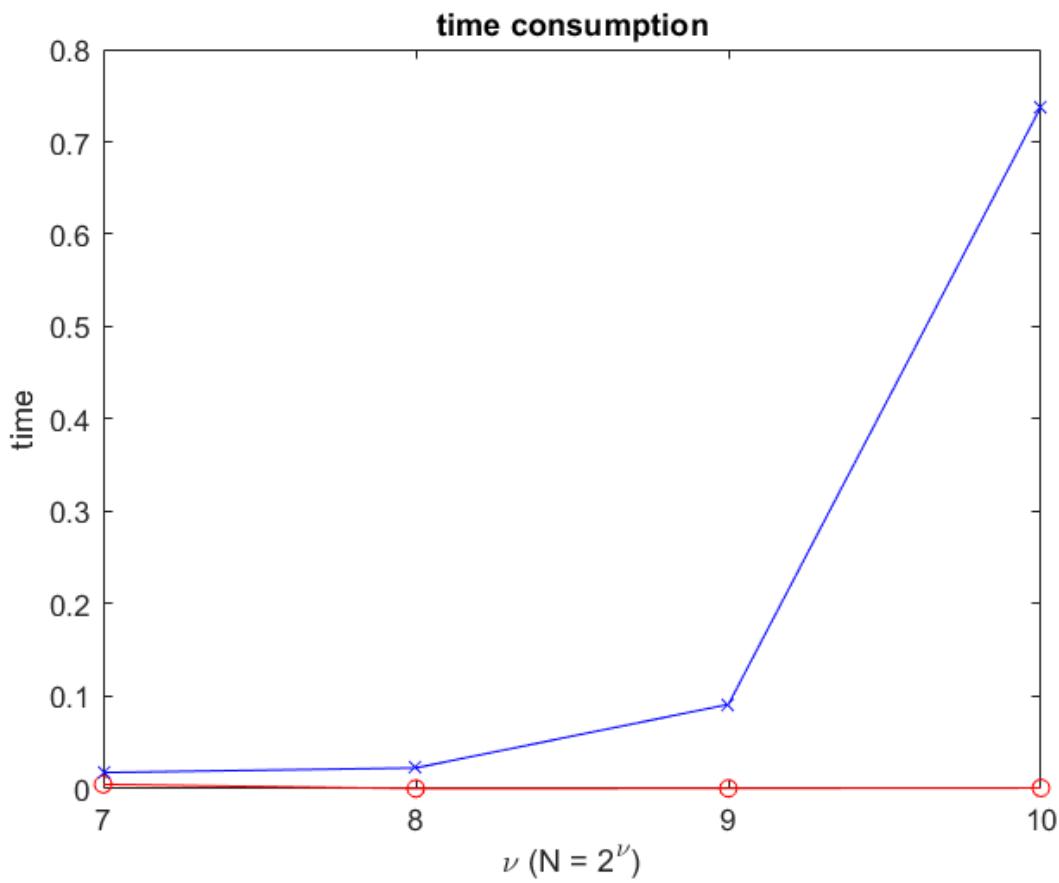
```

% see function dftdirect_m(x,W) at the end of the script
time_dir = [];
time_m = [];
n = [128 256 512 1024];
nu = log2(n);
for N = n
    x = randn(1,N) + 1j*randn(1,N);
    tic;
    X = dftdirect(x);
    time_dir = cat(1,time_dir,toc);
end

for N = n
    x = randn(1,N) + 1j*randn(1,N);
    W = dft_matrix(N);
    tic;
    X = dftdirect_m(x,W);
    time_m = cat(1,time_m,toc);
end

% plot
figure
plot(nu,time_m,'o-','color','red');hold on;
plot(nu,time_dir,'x-','color','blue')
title('time consumption')
xlabel('\nu (N = 2^{\nu})')
ylabel('time')
set(gca, 'XTick', nu)

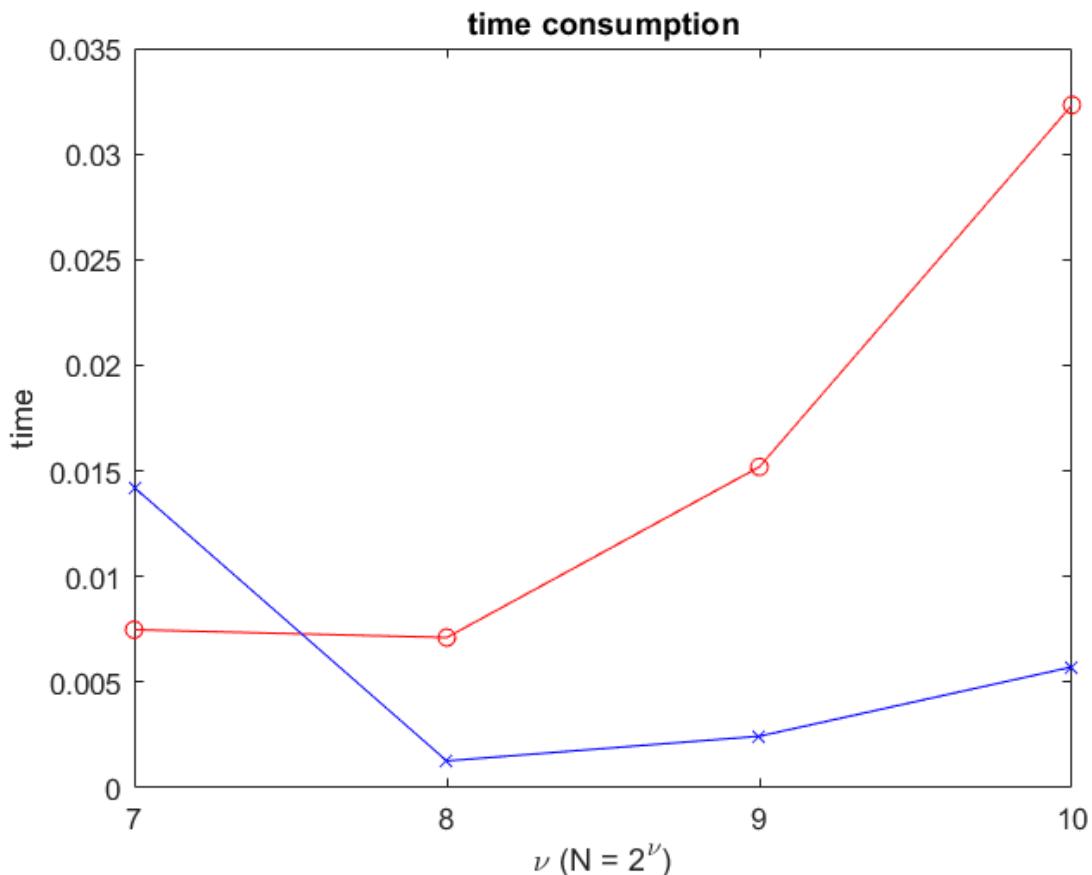
```



(c) (5%) Write a function **X = fftrecur_m(x,W)** that modifies the **fftrecur** function given on page 439 using the matrix W from (a). Using the **tic** and **toc** functions compare computation times for the **fftrecur** and **fftrecur_m** function for $N = 128, 256, 512$, and 1024 . For this purpose generate an N -point complex valued signal as $x = \text{randn}(1,N) + 1j\text{randn}(1,N)$.

```
% see function fftrecur_m(x,W) at the end of the script
time_m = [];
time_dir = [];
n = [128 256 512 1024];
nu = log2(n);
for N = n
    x = randn(1,N) + 1j*randn(1,N);
    W = dft_matrix(N);
    tic;
    X = fftrecur(x);
    time_dir = cat(1,time_dir,toc);
end
for N = n
    x = randn(1,N) + 1j*randn(1,N);
    W = dft_matrix(N);
    tic;
    X = fftrecur_m(x,W);
    time_m = cat(1,time_m,toc);
end
```

```
% plot
figure
plot(nu,time_m,'o-','color','red');hold on;
plot(nu,time_dir,'x-','color','blue')
title('time consumption')
xlabel('\nu (N = 2^{\nu})')
ylabel('time')
set(gca, 'XTick', nu)
```



P3 (15%)

Consider the flow graph in **Figure 8.10** which implements a DIT-FFT algorithm with both input and output in natural order. Let the nodes at each stage be labeled as $sm[k]$, $0 \leq m \leq 3$ with $s0[k] = x[k]$ and $s3[k] = X[k]$, $0 \leq k \leq 7$.

- (a) (5%) Express $sm[k]$ in terms of $sm-1[k]$ for $m = 1, 2, 3$.

$$\begin{cases} s_1[k] = s_0[k] + s_0[k+4]W_8^0 \\ s_1[k+4] = s_0[k] - s_0[k+4]W_8^0 \end{cases} \quad k = 0, 1, 2, 3$$

$$\begin{cases} s_2[k] = s_1[k] + s_1[k+2]W_8^0, \\ s_2[k] = s_1[k+2] + s_1[k+4]W_8^2, \\ s_2[k+4] = s_1[k] - s_1[k+2]W_8^0, \\ s_2[k+4] = s_1[k+2] - s_1[k+4]W_8^2, \end{cases} \quad \begin{matrix} k = 0, 1 \\ k = 2, 3 \\ k = 0, 1 \\ k = 2, 3 \end{matrix}$$

$$\begin{cases} s_3[k] = s_2[2k] + s_2[2k+1]W_8^k \\ s_3[k+4] = s_2[2k] - s_2[2k+1]W_8^k \end{cases} \quad k = 0, 1, 2, 3$$

(b) (5%) Write a MATLAB function $X = \text{fftalt8}(x)$ that computes an 8-point DFT using the equations in part (a). Verify with sequence $x[n] = \{0,1,2,2,2,3,3,3,4\}$.

```
% see function fftalt8(x) at the end of the script  
x = [0 1 2 2 3 3 3 4];  
fft(x)
```

```
ans = 1x8 complex
18.0000 + 0.0000i -3.0000 + 3.8284i -2.0000 + 2.0000i -3.0000 + 1.8284i ...
```

```
X = fftalt8(x)
```

```
X = 1x8 complex
18.0000 + 0.0000i -3.0000 + 3.8284i -2.0000 + 2.0000i -3.0000 + 1.8284i ...
```

(c) (5%) Compare the coding complexity of the above function with that of MATLAB function `fftditr2` shown in **Figure 8.6**, and comment on its usefulness.

ans: The coding complexity of the above function is much larger than that of the `fftditr2` function since the equations are not recursive.

P4 (10%)

Using the flow graph of Figure 8.13 and following the approach used in developing the `fftditr2` function,

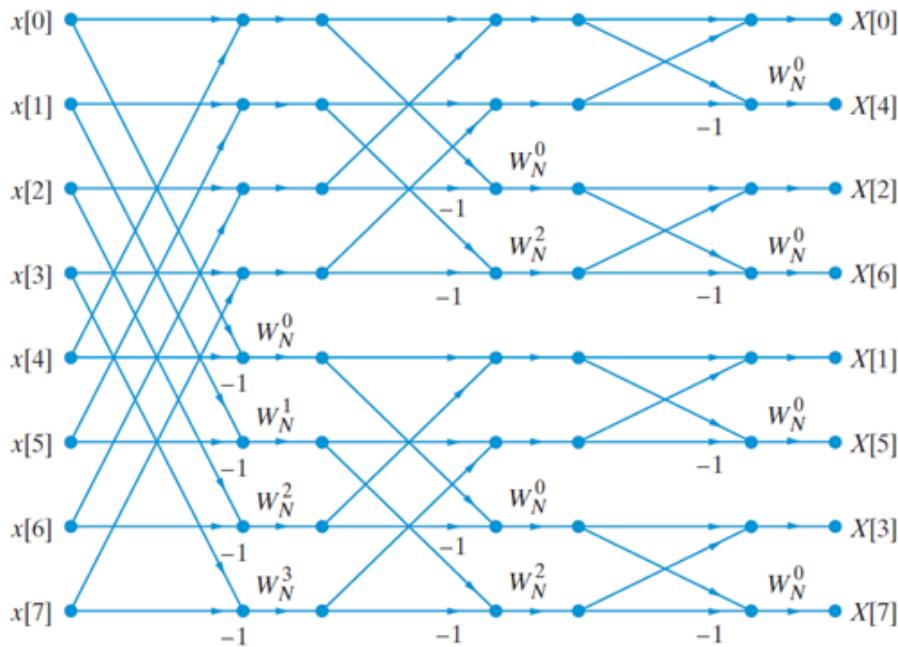


Figure 8.13 Flow graph for the decimation-in-frequency 8-point FFT algorithm. The input sequence is in natural order and the output sequence in bit-reversed order.

(a) (5%) Develop a radix-2 DIF-FFT function $X = \text{fftdifr2}(x)$ for power-of-2 length N .

```
% see function fftdifr2(x) at the end of the script
```

(b) (5%) Verify your function for $N = 2^v$ where $2 \leq v \leq 10$. For this purpose generate an N -point complex-valued signal as $x = \text{randn}(1,N) + 1j\text{randn}(1,N)$.

```

for v = 2:10
    N = 2^v;
    x = randn(1,N) + 1j*randn(1,N);
    X_fft = fft(x);
    X_fftdit = fftdifr2(x);
    diff = abs(X_fft-X_fftdit)>10^-7;
    if(sum(diff)>0)
        display(['error when v= ',num2str(v)]);
    else
        display(['pass for v= ',num2str(v)]);
    end
end

```

```

pass for v= 2
pass for v= 3
pass for v= 4
pass for v= 5
pass for v= 6
pass for v= 7
pass for v= 8
pass for v= 9
pass for v= 10

```

P5(10%)

The **filterfirdf** implements the FIR direct form structure.

(a) (5%)Develop a new MATLAB function **y=filterfirlp(h,x)** that implements the FIR linear-phase form given its impulse response in **h**. This function should first check if **h** is one of type-I through type-IV and then simulate the corresponding equations. If **h** does not correspond to one of the four types then the function should display an appropriate error message.

```
% see function filterfirlp(h,x) below
```

(b) (5%)Verify your function on each of the following FIR systems:

$$h1[n] = \{1,2,3,2,1\},$$

↑

$$h2[n] = \{1,-2,3,3,-2,1\},$$

↑

$$h3[n] = \{1,-5,0,5,-1\},$$

↑

$$h4[n] = \{1,-3,-4,4,3,-1\},$$

↑

$$h5[n] = \{1,2,3,-2,-1\},$$

↑

```
close all; clc
n = 0:9;
xn = ones(size(n));
%% h1:
h = [1 2 3 2 1];
y = filterfirlp(h,xn);
y_ref = filter(h,1,xn);
max(abs(y-y_ref))
```

ans = 0

```
%% h2:
h = [1 -2 3 3 -2 1];
y = filterfirlp(h,xn);
y_ref = filter(h,1,xn);
max(abs(y-y_ref))
```

ans = 0

```
%% h3:
h = [1 -5 0 5 -1];
y = filterfirlp(h,xn);
```

```
y_ref = filter(h,1,xn);
max(abs(y-y_ref))
```

ans = 0

```
%% h4:
h = [1 -3 -4 4 3 -1];
y = filterfirlp(h,xn);
y_ref = filter(h,1,xn);
max(abs(y-y_ref))
```

ans = 0

```
%% h5:
h = [1 2 3 -2 -1];
y = filterfirlp(h,xn);
```

Error using HW6_ans>filterfirlp (line 216)
Impulse Response is not symmetric

P6 (10%)

Consider the IIR normal direct form II structure given in **Figure 9.6** and implemented by (9.18)and (9.20).

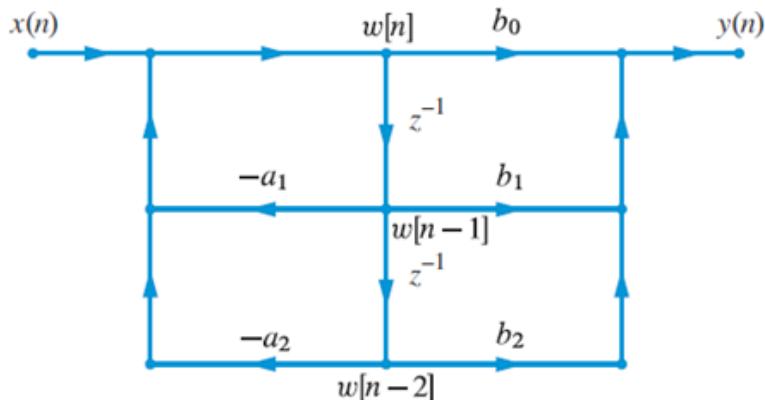


Figure 9.6 Direct form II structure for implementation of an N th order system. For convenience, we assume that $N = M = 2$. If $N \neq M$, some of the coefficients will be zero.

$$y[n] = \sum_{k=0}^{M} b_k w[n-k].$$

(a) (5%) Using the MATLAB function **filterdf1** as a guide, develop a MATLAB function **y=filterdf2(b,a,x)** that implements the normal direct form II structure. Assume zero initial conditions.

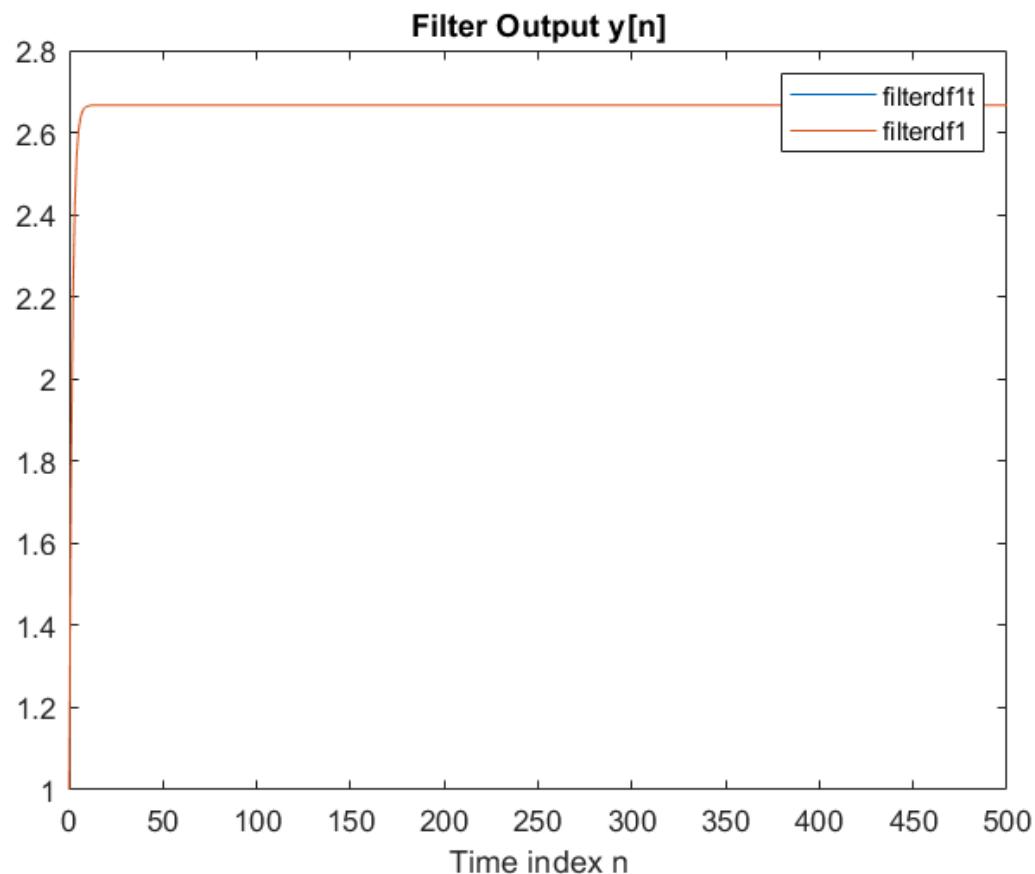
```
% see function filterdf2(b,a,x) below
```

(b) (5%)Determine $y[n]$, $0 \leq n \leq 500$ using your function and filterdf1 function with following inputs:

$$x[n] = \left(\frac{1}{4}\right)^n u[n], a = \begin{bmatrix} 1 & -\frac{3}{2} & \frac{1}{2} \end{bmatrix}, b = 1$$

Compare your results to verify that your filterdf2function is correctly implemented.

```
b = 1;
a = [1 -3/2 1/2];
n = 0:500;
%% Numerical Result 1:
xn = (1/4).^n;
yn = filterdf2(b,a,xn);
yn_ref = filterdf1(b,a,xn);
%% plot:
plot(n,yn,n,yn_ref)
xlabel('Time index n')
title('Filter Output y[n]')
legend('filterdf1t','filterdf1','location','northeast')
```



P7(20%)

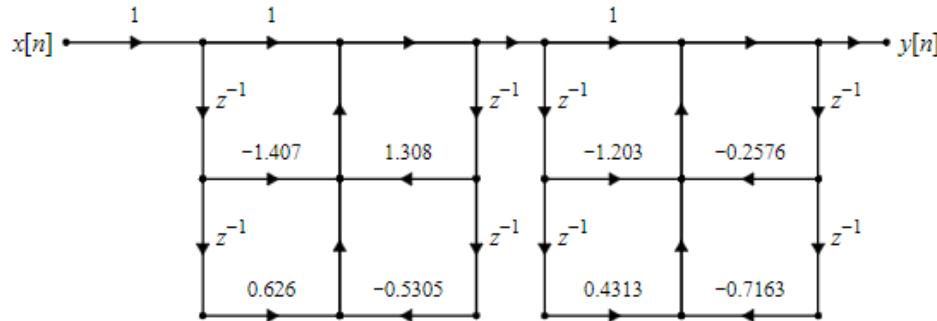
The following numerator and denominator arrays in MATLAB represent the system function of a discrete-time system in direct form:

$$b = [1, -2.61, 2.75, -1.36, 0.27], a = [1, -1.05, 0.91, -0.8, 0.38].$$

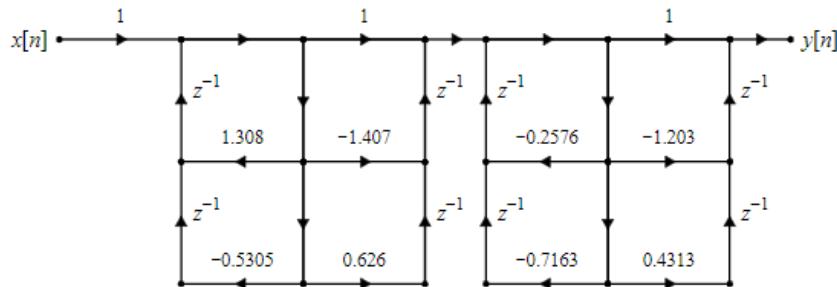
Determine and draw each of the following structures:

```
b = [1, -2.61, 2.75, -1.36, 0.27];
a = [1, -1.05, 0.91, -0.8, 0.38];
tf2sos(b,a)
```

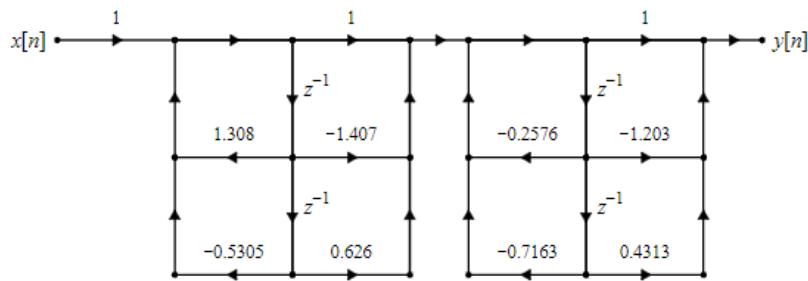
(a) (5%) Cascade form with second-order sections in normal direct form I,



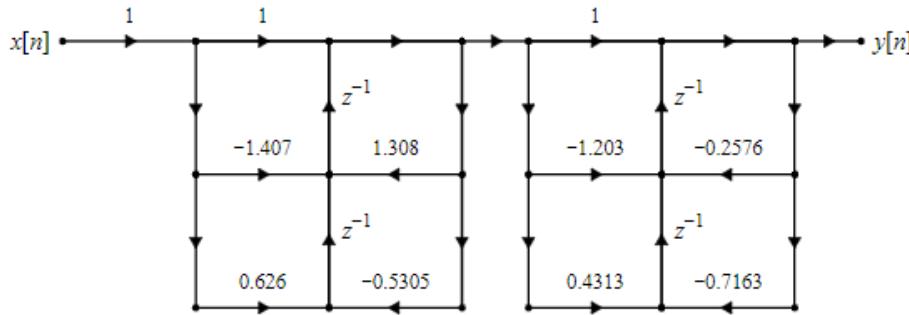
(b) (5%) Cascade form with second-order sections in transposed direct form I,



(c) (5%) Cascade form with second-order sections in normal direct form II,



(d) (5%) Cascade form with second-order sections in transposed direct form II



P8 (10%)

The frequency-sampling form is developed using (9.50) which uses complex arithmetic.

$$H(z) = \frac{1 - z^{-N}}{N} \sum_{k=0}^{N-1} \frac{H[k]}{1 - z^{-1} e^{j \frac{2\pi}{N} k}}, \quad H[k] = H(z) \Big|_{z=e^{j 2\pi k/N}}, \quad (9.50)$$

(a) (5%) Develop a MATLAB function **[G,sos]=firdf2fs(h)** that determines frequency sampling form parameters given in (9.51) and (9.52) given the impulse response in **h**. The matrix **sos** should contain second-order section coefficients in the form similar to the **tf2sos** function while **G** array should contain the respective gains of second-order sections. Incorporate the coefficients for the $H[0]$ and $H[N/2]$ terms in **sos** and **G** arrays.

```
% see function firdf2fs(h) below
```

(b) (2%) Verify your function by input **h** with sampled frequency response (9.53) and compare with the system function (9.54)

$$H[k] = H(e^{j \frac{2\pi}{33} k}) = e^{-j \frac{32\pi}{33} k} \times \begin{cases} 1, & k = 0, 1, 2, 31, 32 \\ 0.5, & k = 3, 30 \\ 0, & \text{otherwise} \end{cases} \quad (9.53)$$

$$\begin{aligned} H(z) = & \frac{1 - z^{-33}}{33} \left[\frac{1}{1 - z^{-1}} + \frac{-1.99 + 1.99z^{-1}}{1 - 1.964z^{-1} + z^{-2}} \right. \\ & \left. + \frac{1.964 - 1.964z^{-1}}{1 - 1.857z^{-1} + z^{-2}} + \frac{-1.96 + 1.96z^{-1}}{1 - 1.683z^{-1} + z^{-2}} \right]. \end{aligned} \quad (9.54)$$

```
N = 33;
alpha = (N-1)/2;
k = 0:N-1;
magHk = [1,1,1,0.5,zeros(1,26),0.5,1,1];
```

```

angHk = -32*pi*k/33;
H = magHk.*exp(1j*angHk);
h = real(ifft(H,N));
[G,sos] = firdf2fs(h)

```

```
[ [G G G].*sos(:,1:3) sos(:,4:end) ]
```

```

ans = 18x6
    1.0000      0      0    1.0000   -1.0000      0
            0      0      0    1.0000   1.0000      0
-1.9909    1.9909      0    1.0000   -1.9639   1.0000
  1.9639   -1.9639      0    1.0000   -1.8567   1.0000
-0.9595    0.9595      0    1.0000   -1.6825   1.0000
  0.0000    0.0000      0    1.0000   -1.4475   1.0000
  0.0000   -0.0000      0    1.0000   -1.1601   1.0000
-0.0000    0.0000      0    1.0000   -0.8308   1.0000
-0.0000    0.0000      0    1.0000   -0.4715   1.0000
-0.0000   -0.0000      0    1.0000   -0.0952   1.0000
    .
    .

```

```

clear
H=zeros(1,33);
H(1)=exp(-j*32*pi/33*0);
H(2)=exp(-j*32*pi/33*1);
H(3)=exp(-j*32*pi/33*2);
H(32)=exp(-j*32*pi/33*31);

```

```
H(33)=exp(-j*32*pi/33*32);
H(4)=exp(-j*32*pi/33*3)*0.5;
H(31)=exp(-j*32*pi/33*30)*0.5;
ifft(H)
```

```
ans = 1x33 complex
0.0004 - 0.0000i 0.0033 - 0.0000i 0.0073 - 0.0000i 0.0095 - 0.0000i ...
```

```
fft(ifft(H))
```

```
ans = 1x33 complex
1.0000 + 0.0000i -0.9955 - 0.0951i 0.9819 + 0.1893i -0.4797 - 0.1409i ...
```

Functions

```
function x = IDFT(X,N)
    X = [X(1) X(end:-1:2)];
    x = fft(X)/N;
end

function D = dft_matrix(n)
    f = 2*pi/n; % Angular increment.
    w = (0:f:2*pi-f/2).'* 1i; % Column.
    x = 0:n-1; % Row.
    D = exp(-w*x); % Exponentiation of outer product.
end

function X = dftdirect_m(x,w)
    X = x*w;
end

function Xdft = fftrecur_m(x,w)
    N = length(x);
    if N ==1
        Xdft = x;
    else
        m = N/2;
        W_2 = w(1:2:N,1:m);
        XE = fftrecur_m(x(1:2:N),W_2);
        XO = fftrecur_m(x(2:2:N),W_2);
        temp = W(1:m,2).*XO;
        Xdft = [ XE+temp ; XE-temp ];
    end
end

function X = fftalt8(x)
    N = 8;
    s = x;
    w = exp(-j*2*pi/N).^(0:N-1);
    % Stage I:
```

```

temp = s;
s(1:4) = temp(1:4)+temp(5:8);
s(5:8) = temp(1:4)-temp(5:8);
temp = s;
s(1:2) = temp(1:2)+temp(3:4);
s(3:4) = temp(5:6)+temp(7:8)*w(3);
s(5:6) = temp(1:2)-temp(3:4);
s(7:8) = temp(5:6)-temp(7:8)*w(3);
% Stage III:
temp = s;
s(1:4) = temp(1:2:end)+temp(2:2:end).*w(1:4);
s(5:8) = temp(1:2:end)-temp(2:2:end).*w(1:4);
X = s;
end

function x=fftdiffr2(x)
N=length(x); nu=log2(N);
for m=nu:-1:1;
    L=2^m;
    L2=L/2;
    for ir=1:L2;
        W=exp(-1i*2*pi*(ir-1)/L);
        for it=ir:L:N;
            ib=it+L2;
            temp=x(it)+x(ib);
            x(ib)=x(it)-x(ib);
            x(ib)=x(ib)*W;
            x(it)=temp;
        end
    end
end
x = bitrevorder(x);
end

function y = filterfirlp(h,x)
nh = length(h);
M = nh-1;
nx = length(x);
x = [zeros(1,M) x(:)'];
y = zeros(1,nx);
eo = mod(M,2) ~= 0;
if max(abs(h + fliplr(h))) == 0
    syasy = 1;
elseif max(abs(h - fliplr(h))) == 0
    syasy = 0;
else
    error('Impulse Response is not symmetric')
end
h = h';
caseind = 2*syasy + eo;
switch caseind
    case 0
        MM = M/2;

```

```

        for n = 1:nx
            y(n) = (x(n+M:-1:n+M-MM+1)+x(n:1:n+MM-1))*h(1:MM)+ h(MM+1)*x(n+M-MM);
        end
    case 1
        MM = (M-1)/2+1;
        for n = 1:nx
            y(n) = (x(n+M:-1:n+M-MM+1)+x(n:1:n+MM-1))*h(1:MM);
        end
    case 2
        MM = M/2;
        for n = 1:nx
            y(n) = (x(n+M:-1:n+M-MM+1)-x(n:1:n+MM-1))*h(1:MM);
        end
    case 3
        MM = (M-1)/2+1;
        for n = 1:nx
            y(n) = (x(n+M:-1:n+M-MM+1)-x(n:1:n+MM-1))*h(1:MM);
        end
    end
end

function [y] = filterdf2(b,a,x)
M = length(b)-1; N = length(a)-1; K = max(M,N);
a0 = a(1); a = reshape(a,1,N+1)/a0;
b = reshape(b,1,M+1)/a0; a = a(2:end);
Lx = length(x); x = [zeros(K,1);x(:)];
Ly = Lx+K; y = zeros(1,Ly); w = zeros(Ly,1);
for n = K+1:Ly
    w(n) = x(n) - a*w(n-1:-1:n-N);
    y(n) = b*w(n:-1:n-M);
end
y = y(K+1:Ly);
end

function [G,sos] = firdf2fs(h)
N = length(h);
if mod(N,2) == 0
    K = N/2-1;
else
    K = (N-1)/2;
end
G = zeros(K+2,1);
H = fft(h);
Hmag = abs(H);
Hang = angle(H);
G(1) = H(1);
G(3:end) = 2*Hmag(2:1+K);
sos = zeros(K+2,6);
sos(1,:) = [1 0 0 1 -1 0];
sos(2,:) = [1 0 0 1 1 0];
for ii = 1:K
    sos(2+ii,:) = [cos(Hang(ii+1)) -cos(Hang(ii+1)-2*pi*ii/N) 0 1 -2*cos(2*pi*ii/N) 1];
end

```

```
if mod(N,2) == 0
    G(2) = H(N/2+1);
else
    G(2) = 0;
end
end
```