

Homework Assignment #2: Chap. 2-4

Due: March 26, 2020

I Program Assignment (100%)

1. (8%) Let $x[n] = h[n] = (0.9)^n u[n]$ and $y[n] = x[n] * h[n]$
 - (a) Determine $y[n]$ analytically and plot the first 99 **non-zero** samples using the `stem` function.
 - (b) Take first 50 samples of $x[n]$ and $h[n]$. Compute and plot $y[n]$ using the `conv` function.
 - (c) Using the `filter` function, determine and plot the first 99 samples of $y[n]$.
 - (d) Which of the outputs in (b) and (c) come close to that in (a)? Explain.

2. (8%) The sum $A_x \triangleq \sum_n x[n]$ can be thought of as a measure of the “area” under a sequence $x[n]$.
 - (a) Starting with the convolution sum (2.36), show that $A_y = A_x A_h$ (derive in the live script).
 - (b) Given the sequences
$$x = \sin(2 * \pi * 0.01 * (0:100)) + 0.05 * \text{randn}(1,101); h = \text{ones}(1,5);$$
compute $y[n] = h[n] * x[n]$, check whether $A_y = A_x A_h$, and use the `subplot` function plot $x[n]$ and $y[n]$ on the same graph.
 - (c) Normalize $h[n]$ so that $A_h = 1$ and repeat part (b).
 - (d) If $A_h = 1$, then $A_y = A_x$. Use this result to explain the difference between the plots obtained in parts (b) and (c).

3. (8%) The response of a LTI system to the input $x[n] = u[n]$ is $y[n] = 2 \cdot (\frac{1}{3})^n u[n]$.
 - (a) Find the impulse response $h[n]$ of the system, and check the results using the function `filter`.
 - (b) Plot the pole-zero pattern using the function `zplane(b, a)`.
 - (c) Compute and plot the impulse response using the functions `filter` and `stem`. Compare with the plot obtained using the function `impz`.
 - (d) Use the function `residuez` and the z -transform pairs in Table 3.1 to find an analytical expression for the impulse response $h[n]$.

4. (12%) Find the impulse response of the system (3.97) for the case of **real and equal poles** and use the result to determine how the location of the poles affects (a) the stability of the system, and (b) the shape of the impulse response. (Hint: Use MATLAB to replicate Figure 3.10 for a double pole, find C and discuss stability of the three cases $|a_1| < C$, $|a_1| = C$, $|a_1| > C$)

5. (10%) Consider the following LCCDE:

$$y[n] = 2\cos(\omega_0)y[n-1] - y[n-2],$$

with no input but with initial conditions $y[-1] = 0$ and $y[-2] = -A\sin(\omega_0)$.

- (a) Show that the solution of the above LCCDE is given by the sequence

$$y[n] = A \cdot \sin[(n+1)\omega_0] \cdot u[n].$$

This system is known as a *digital oscillator* (derive in the live script).

- (b) For $A = 2$ and $\omega_0 = 0.1\pi$, verify the operation of the above digital oscillator using the `filtic` and the `filter` function. (Hint: Check one-sided z-transform in supplement)

6. (10%) In this problem we illustrate the numerical evaluation of DTFS using MATLAB.

- (a) Write a function `c=dtfs0(x)` which computes the DTFS coefficients (4.67) of a periodic signal and verify the result with `dtfs` function.
- (b) Write a function `x=idtfs0(c)` which computes the inverse DTFS (4.63) and verify the result with `idtfs` function.

7. (12%) Determine and plot the magnitude and phase spectra of the following periodic sequences:

(a) $x_1[n] = 4 \cdot \cos(1.2\pi n + \frac{\pi}{3}) + 6 \cdot \sin(0.4\pi n - \frac{\pi}{6})$

(b) $x_2[n] = \{1, 1, 0, 1, 1, 1, 0, 1\}$, (one period)

(c) $x_3[n] = 1 - \sin(\frac{\pi}{4}n)$, $0 \leq n \leq 11$ (one period)

8. (8%) Consider a noncausal finite length sequence $x[n] = \delta[n+1] + \delta[n] + \delta[n-1]$, we can compute the DTFT in MATLAB using following scripts:

```
x=[1 1 1]; % n=-1,0,1
om=linspace(-pi, pi, 60);
X1=dtft12(x, -1, om); X2=freqz(x, 1, om);
```

- (a) Use `subplot` to plot the magnitude $|X1|, |X2|$ and phase $\angle X1, \angle X2$ in one graph respectively.
- (b) Is the magnitude $|X1| = |X2|$? Is the phase $\angle X1 = \angle X2$? If not, Explain.

9. (12%) In this problem use the image file “DSP.png” which has 100×300 pixels (unsigned 8 bits per pixel). You can use the following MATLAB script to load, show and store the image files:

```
img = imread('DSP.png');
imshow(img);
imwrite(img, 'DSP0.png');
```

- (a) Consider the 5×5 impulse response $h[m, n]$ given as follow:

$$h[m, n] = \begin{cases} \frac{1}{25}, & -2 \leq m, n \leq 2 \\ 0, & \text{otherwise} \end{cases},$$

filter the “DSP.png” using (2.78) and display the resulting image, comment on the result. (Hint: You can directly use `conv2` function, and make sure the data format is `double` before filtering).

- (b) Repeat part (a) and try two different kernels $h_1[m, n]$ and $h_2[m, n]$: (known as Sobel filter)

$$h_1[m, n] = \begin{bmatrix} 1 & 0 & -1 \\ 2 & 0 & -2 \\ 1 & 0 & -1 \end{bmatrix}, \quad h_2[m, n] = \begin{bmatrix} 1 & 2 & 1 \\ 0 & 0 & 0 \\ -1 & -2 & -1 \end{bmatrix},$$

what is the difference between the two results? (Hint: You should do `abs` and `uint8` before `imshow`)

- (c) Repeat (b) by using `filter2` function, what is the difference between the two functions? (Hint: Check the pixel values before `abs`)

10. (12%) This problem uses the sound file “handel.wav” available in MATLAB. This sound is sampled at $F_s = 8192$ samples per second using 8-bits per sample. You can use the following MATLAB script to load, play and store the audio files:

```
[y,Fs] = audioread('handel.wav');
playerObj = audioplayer(y, Fs);
play(playerObj);
audiowrite('handel0.wav', y, Fs);
```

- (a) Select every other sample in audio signal y which reduces the sampling rate by a factor of two. Now listen to the new sound array using the sound function at half the sampling rate.

(Hint: You can use the logical array to index array elements)

- (b) Select every fourth sample in audio signal y which reduces the sampling rate by a factor of four. Listen to the resulting sound array using the sound function at quarter the sampling rate.

- (c) From the results of (a) and (b), what do you discover?

II Reference

1. Formula 2.36

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]. \quad -\infty < n < \infty \quad (2.36)$$

2. Formula 3.97

$$H(z) = \frac{b_0 + b_1z^{-1}}{1 + a_1z^{-1} + a_2z^{-2}} = \frac{z(b_0z + b_1)}{z^2 + a_1z + a_2}. \quad (3.97)$$

3. Figure 3.10

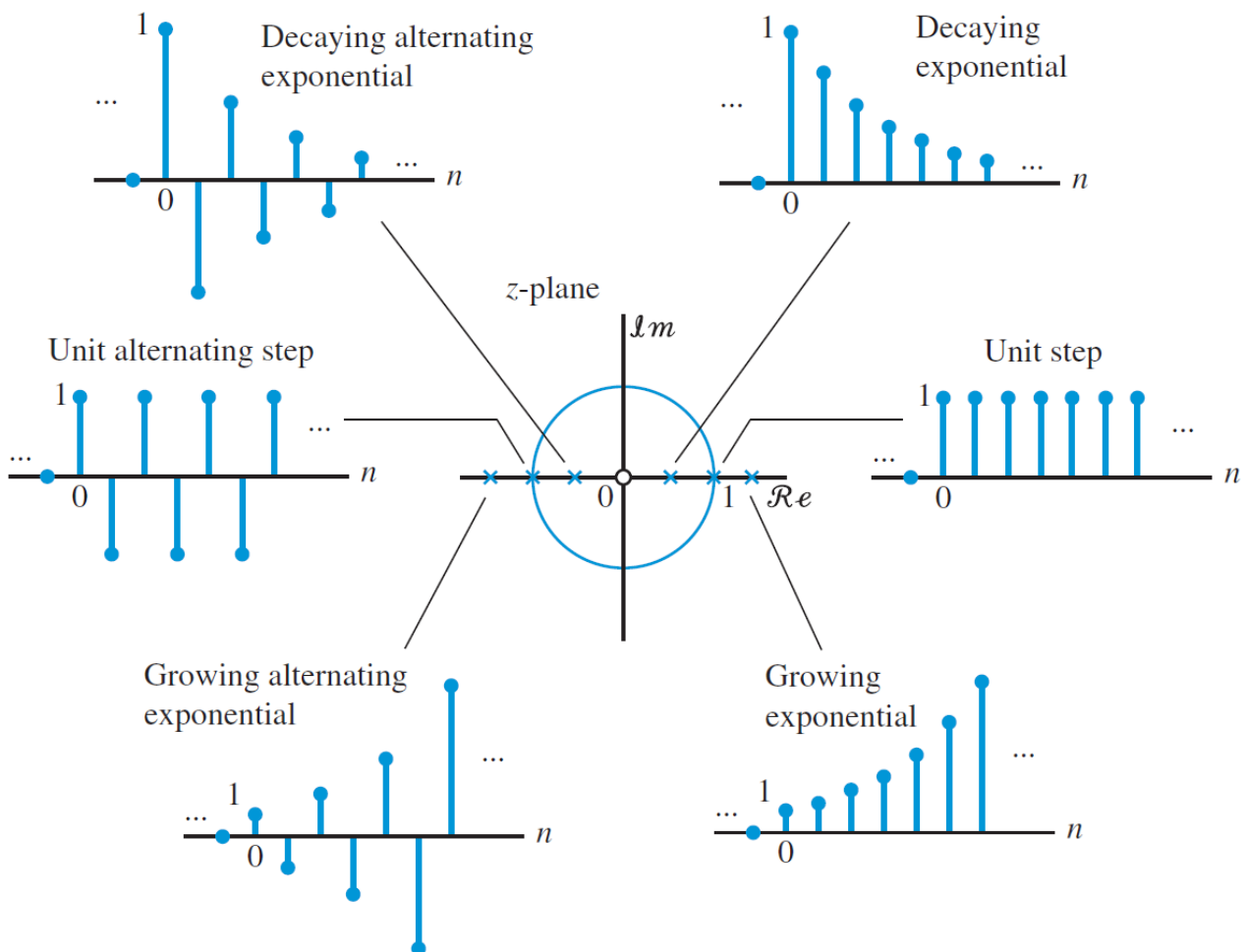


Figure 3.10 Impulse responses associated with real poles in the z -plane. Only the two poles inside the unit circle correspond to stable systems.

4. Table 3.1

Table 3.1 Some common z-transform pairs			
	Sequence $x[n]$	z-Transform $X(z)$	ROC
1.	$\delta[n]$	1	All z
2.	$u[n]$	$\frac{1}{1 - z^{-1}}$	$ z > 1$
3.	$a^n u[n]$	$\frac{1}{1 - az^{-1}}$	$ z > a $
4.	$-a^n u[-n - 1]$	$\frac{1}{1 - az^{-1}}$	$ z < a $
5.	$na^n u[n]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z > a $
6.	$-na^n u[-n - 1]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z < a $
7.	$(\cos \omega_0 n)u[n]$	$\frac{1 - (\cos \omega_0)z^{-1}}{1 - 2(\cos \omega_0)z^{-1} + z^{-2}}$	$ z > 1$
8.	$(\sin \omega_0 n)u[n]$	$\frac{(\sin \omega_0)z^{-1}}{1 - 2(\cos \omega_0)z^{-1} + z^{-2}}$	$ z > 1$
9.	$(r^n \cos \omega_0 n)u[n]$	$\frac{1 - (r \cos \omega_0)z^{-1}}{1 - 2(r \cos \omega_0)z^{-1} + r^2 z^{-2}}$	$ z > r$
10.	$(r^n \sin \omega_0 n)u[n]$	$\frac{(\sin \omega_0)z^{-1}}{1 - 2(r \cos \omega_0)z^{-1} + r^2 z^{-2}}$	$ z > r$

5. Formula 4.67

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} kn}. \quad (4.67)$$

6. Formula 4.63

$$x[n] = \sum_{k=0}^{N-1} c_k e^{j \frac{2\pi}{N} kn}. \quad (4.63)$$

7. Formula 2.78

$$y[m, n] = \sum_{k=m-K}^{m+K} \sum_{\ell=n-L}^{n+L} x[k, \ell] h[m - k, n - \ell]. \quad (2.78)$$

III Supplement

- **One-Sided Z Transform** (It is not included in this course, but important in realistic discrete system)

In some practical case (only have one-sided sequence $x[n]$ and initial condition), our solutions require the *one-sided* or *unilateral z-transform*, defined by the formula:

$$X^+(z) \triangleq \mathcal{Z}^+\{x[n]\} \triangleq \sum_{n=0}^{\infty} x[n]z^{-n}. \quad (3.100)$$

The one-sided z-transform has the following characteristics:

1. It does not contain information about the signal $x[n]$ for negative values of time (i.e., for $n < 0$)
2. It is unique only for causal signals, because only these signals are zero for $n < 0$.
3. The one-sided z-transform $X^+(z)$ of $x[n]$ is identical to the two-sided z-transform $X(z)$ of the signal $x[n]u[n]$.
4. ROC of $X^+(z)$ is always the exterior of the circle, so it is not necessary to refer to their ROC.

(More details in the textbook, chapter 3.8)

✘ One-Sided Z-Transform of Two-Order LCCDE

We can define two-order **linear constant-coefficient difference equation (LCCDE)** as follows:

$$y[n] = b_0x[n] + b_1x[n-1] + b_2x[n-2] - a_1y[n-1] - a_2y[n-2],$$

Apply one-sided z-transform, we have

$$Y^+(z) = \frac{b_0 + b_1z^{-1} + b_2z^{-2}}{1 + a_1z^{-1} + a_2z^{-2}} X^+(z) + \frac{(b_2x[-1] - a_2y[-1])z^{-1} + (b_1x[-1] + b_2x[-2] - a_1y[-1] - a_2y[-2])}{1 + a_1z^{-1} + a_2z^{-2}},$$

(You can directly use this equation to derive problem 5(a))

✘ MATLAB Usage (Reference)

```
b = [b0 b1 b2];  
a = [1 a1 a2];  
yic = [y[-1] y[-2]];  
xic = [x[-1] x[-2]];  
zic = filtic(b, a, yic, xic);  
y = filter(b, a, x, zic);
```

(You can reference the scripts to verify the operation in problem 5(b))

(More details in the textbook, page 120)