National Tsing Hua University Department of Electrical Engineering EE3660 Intro. to Digital Signal Processing, Spring 2020

Homework Assignment #1: Chap. 2-4 Solution

I Paper Assignment (100%)

1. (12%) Determine whether the following systems are linear, time-invariant, causal, and stable.

(a) y[n] = x[-n](b) $y[n] = \cos(\pi n)x[n]$ (c) $y[n] = \sum_{k=n-1}^{k=\infty} x[k]$

Solution

(a) y[n] = x[-n]

(i) Linear :

y[n] = L(x[n]) = x[-n],Let $y3[n] = L(a^*x1[n] + b^*x2[n])$ $= a^*x1[-n] + b^*x2[-n]$ $= a^*L(x1[n]) + b^*L(x2[n])$ $= a^*y1[n] + b^*y2[n]$ It's a linear system.

(ii) Time-invariant :

This is a "flip" function. L(x[n - n0]) = x[-n - n0] $\neq y[n - n0] = x[-(n - n0)] = x[-n + n0]$ It's not a time-invariant system. (iii) Causal :

When n < 0, the values of y[n] depend on the future values of x[n]It's not a causal system.

(iv) Stable :

 $|y[n]| \le |x[-n]| \le M$, It's a stable system.

(b) $y[n] = cos(\pi n)x[n]$

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(i) Linear :
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y[n] = L(x[n]) = \cos(\pi n)x[n],

Let y3[n] = L(a^*x1[n] + b^*x2[n])

= \cos(\pi n)(a^*x1[n] + b^*x2[n])

= a^*\cos(\pi n)x1[n] + b^*\cos(\pi n)x2[n]

= a^*L(x1[n]) + b^*L(x2[n])

= a^*y1[n] + b^*y2[n]

It's a linear system.
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(ii) Time-invariant :
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 $y[n] = L(x[n]) = (-1)^{n} x[n],$ $L(x[n-1]) = (-1)^{n} x[n-1]$ $\neq y[n-1] = (-1)^{n-1} x[n-1]$ It's not a time-invariant system.

(iii) Causal:

It's each output only depends on the current value of x[n]. It's a causal system.

(iv) Stable :

 $\cos(\pi n) = (-1)^n, \ n \in \mathbb{Z}$

This system only changes the sign of x[n], and doesn't change the magnitude of it.

It's a stable system.

(c)
$$y[n] = \sum_{k=n-1}^{k=\infty} x[k]$$

(i) Linear :

$$y[n] = L(x[n]) = \sum_{k=n-1}^{k=\infty} x[k]$$

Let $y3[n] = L(a^*x1[n] + b^*x2[n])$

$$= \sum_{k=n-1}^{k=\infty} (a^*x1[k] + b^*x2[k])$$

$$= a^* \sum_{k=n-1}^{k=\infty} x1[k] + b^* \sum_{k=n-1}^{k=\infty} x2[k]$$

$$= a^*L(x1[n]) + b^*L(x2[n])$$

$$= a^*y1[n] + b^*y2[n]$$

It's a linear system.

(ii) Time-invariant :

$$y[n] = L(x[n]) = \sum_{k=n-1}^{k=\infty} x[k],$$
$$L(x[n-n0]) = \sum_{k=n-n0-1}^{k=\infty} x[k],$$
$$= y[n-n0]$$
It's a time-invariant system.

(iii) Causal:

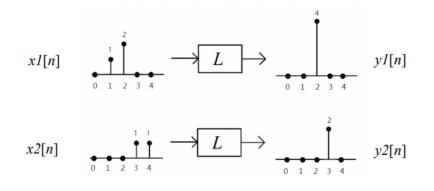
It's not causal, since it sums forward in time. It's not a causal system.

(iv) Stable :

Let
$$\mathbf{x}[\mathbf{n}] = \delta[\mathbf{n}], \ \sum_{k=n-1}^{k=\infty} \delta[k] = \infty$$

It's not a stable system.

2. (10%) A linear system L generates output signals: y1[n], y2[n] in response to the input signals xl[n], x2[n] respectively. Determine whether the system L is time invariant.



Solution

Suppose system L is time invariant,

 $\delta[n] = x1[n+2] - x2[n+4] = \{1, 0, 0, 0, 0\}$ \uparrow Due to L is linear, $h[n] = y1[n+2] - y2[n+4] = \{-2, 4, 0, 0, 0\}$ \uparrow $x1[n] = \delta[n-1] + 2*\delta[n-2]$ $j1[n] = h[n-1] + 2*h[n-2] = \{0, -2, 4, 0, 0\} + \{0, 0, -4, 8, 0\}$ \uparrow $= \{0, -2, 0, 8, 0\} \neq y1[n] = \{0, 0, 4, 0, 0\}$ \uparrow

L is not time invariant system.

3. (12%) Given the z-transform pair $x[n] \leftrightarrow X(z) = \frac{1}{(1-2z^{-1})}$ with ROC: |z| < 2, use the z-transform properties to determine the z-transform of the following sequences:

Solution

(a)
$$y[n] = (\frac{1}{3})^n x[n]$$

Scaling :

$$Y(z) = X(3z) = \frac{1}{1 - \frac{2}{3}z^{-1}}, \text{ ROC} : |z| < \frac{2}{3}$$

(a)
$$y[n] = x[n] * x[-n]$$

Folding and convolution :

$$Y(z) = X(z)X\left(\frac{1}{z}\right) = \frac{1}{1-2z^{-1}}\frac{1}{1-2z} = \frac{-1/2}{1-\frac{5}{2}z^{-1}+z^{-2}} , \text{ ROC} : 0.5 < |z| < 2$$

(b) y[n] = nx[n]

Differentiation :

$$Y(z) = -z \frac{dX(z)}{dz} = \frac{2z^{-1}}{1 - 4z^{-1} + 4z^{-2}}$$
, ROC : $|z| < 2$

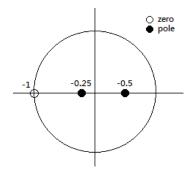
4. (12%) A causal LTI system has impulse response h[n], for which the z-transform is

$$H(z) = \frac{1 + z^{-1}}{(1 - 0.5z^{-1})(1 + 0.25z^{-1})}$$

- (a) Draw the pole-zero plot of H(z) and specify its ROC.
- (b) Explain whether the system is stable?
- (c) Find the impulse response h[*n*] of the system.

Solution

(a) Draw the pole-zero plot of H(z) and specify its ROC.



Due to H(z) is a causal system, ROC should include $|z| = \infty$ ROC : |z| > 0.5

(b) Explain whether the system is stable?

Due to the ROC includes the unit circle, this system is stable.

(c) Find the impulse response h[*n*] of the system.

$$H(z) = \frac{1+z^{-1}}{(1-0.5z^{-1})(1+0.25z^{-1})} = \frac{2}{(1-0.5z^{-1})} - \frac{1}{(1+0.25z^{-1})}$$
$$h[n] = 2\left(\frac{1}{2}\right)^n u[n] - \left(-\frac{1}{4}\right)^n u[n]$$

5. (15%) Use the method of partial fraction expansion to determine the sequences corresponding to the following z-transforms:

(a)
$$X(z) = \frac{z}{z^3 + 2z^2 + \frac{5}{4}z + \frac{1}{4}}$$
, $|z| > 1$.
(b) $X(z) = \frac{z}{(z^2 - \frac{1}{3})^2}$, $|z| < 0.5$.

Solution

(a)
$$X(z) = \frac{z}{z^{3} + 2z^{2} + \frac{5}{4}z + \frac{1}{4}}$$
, $|z| > 1$.
 $X(z) = \frac{z^{-2}}{(1+z^{-1})(1+0.5z^{-1})^{2}}$,
 $X(z) = \frac{A}{1+z^{-1}} + \frac{B}{1+0.5z^{-1}} + \frac{C}{(1+0.5z^{-1})^{2}}$,
 $A = (1 + z^{-1})X(z)|_{z^{-1}=-1} = 4$,
 $C = (1 + 0.5z^{-1})^{2}X(z)|_{z^{-1}=-0.5} = -4$,
 $z^{-2} = (1 + 0.5z^{-1})^{2}A + (1 + z^{-1})(1 + 0.5z^{-1})B + (1 + z^{-1})C$
 $z^{-2} = (1 + 0.5z^{-1})^{2}A + (1 + z^{-1})(1 + 0.5z^{-1})B - (1 + z^{-1})A$
 $z^{-2} = z^{-2} + 0.5z^{-2}B$, $B = 0$, (focus on z^{-2} coefficients)
 $X(z) = \frac{4}{1+z^{-1}} + \frac{-4}{(1+0.5z^{-1})^{2}}$
 $X(z) = \frac{4}{1+z^{-1}} + \frac{8(\cdot 0.5)z^{-1}}{(1+0.5z^{-1})^{2}}z$
 $na^{n}u[n] \frac{az^{-1}}{(1 - az^{-1})^{2}}$
 $x[n] = 4(-1)^{n}u[n] + 8(n + 1)(-0.5)^{n+1}u[n + 1]$
 $x[n] = 4(-1)^{n}u[n] - 4(n + 1)(-0.5)^{n}u[n]$
 $x[n] = 4(-1)^{n}u[n] - 4(n - 0.5)^{n}u[n]$
(b) $X(z) = \frac{z}{(z^{-1}-\frac{1}{3})^{2}}$, $|z| < 0.5$.
 $X(z) = \frac{z}{(z^{-1}-\frac{1}{3})^{2}(z+\frac{1}{\sqrt{3}})^{2}} = \frac{A}{z-\frac{1}{\sqrt{3}}} + \frac{B}{(z-\frac{1}{\sqrt{3}})^{2}} + \frac{C}{z+\frac{1}{\sqrt{3}}} + \frac{D}{(z+\frac{1}{\sqrt{3}})^{2}}$,

$$B = (z - \frac{1}{\sqrt{3}})^2 X(z)|_{z = \frac{1}{\sqrt{3}}} = \frac{\sqrt{3}}{4},$$

$$D = (z + \frac{1}{\sqrt{3}})^2 X(z)|_{z = -\frac{1}{\sqrt{3}}} = \frac{-\sqrt{3}}{4},$$

$$z = (z - \frac{1}{\sqrt{3}})(z + \frac{1}{\sqrt{3}})^2 A + (z + \frac{1}{\sqrt{3}})^2 B + (z - \frac{1}{\sqrt{3}})^2 (z + \frac{1}{\sqrt{3}}) C + (z - \frac{1}{\sqrt{3}})^2 D$$

$$0 = \frac{-1}{3} \frac{1}{\sqrt{3}} A + \frac{1}{3} \frac{\sqrt{3}}{4} + \frac{1}{3} \frac{1}{\sqrt{3}} C + \frac{1}{3} \frac{-\sqrt{3}}{4}, A = C, \text{ (focus on constant coefficients)}$$

$$0 = \frac{1}{\sqrt{3}} z^2 A + z^2 \frac{\sqrt{3}}{4} + \frac{1}{\sqrt{3}} z^2 C + z^2 \frac{-\sqrt{3}}{4}, A = -C, \text{ (focus on } z^2 \text{ coefficients)}$$

$$A = C = 0,$$

$$X(z) = \frac{\frac{\sqrt{3}}{4}}{(z - \frac{1}{\sqrt{3}})^2} + \frac{\frac{-\sqrt{3}}{4}}{(z + \frac{1}{\sqrt{3}})^2},$$

$$X(z) = \frac{\frac{3}{4} \frac{1}{\sqrt{3}} z^{-1}}{(1 - \frac{1}{\sqrt{3}} z^{-1})^2} z^{-1} + \frac{\frac{3}{4} \frac{-1}{\sqrt{3}} z^{-1}}{(1 + \frac{1}{\sqrt{3}} z^{-1})^2} z^{-1}, \quad -nd^n u[-n-1] \quad \frac{az^{-1}}{(1 - az^{-1})^2}$$

$$\begin{aligned} \mathbf{x}[\mathbf{n}] &= \frac{-3}{4} (n-1) (\frac{1}{\sqrt{3}})^{n-1} u[-(n-1)-1] + \frac{-3}{4} (n-1) (\frac{-1}{\sqrt{3}})^{n-1} u[-(n-1)-1] \\ \mathbf{x}[\mathbf{n}] &= \frac{-3}{4} (n-1) (\frac{1}{\sqrt{3}})^{n-1} u[-n] - \frac{3}{4} (n-1) (\frac{-1}{\sqrt{3}})^{n-1} u[-n] \end{aligned}$$

6. (12%) A function called autocorrelation for a real-valued, absolutely summable sequence x[n], is defined as

$$r_{xx}[\ell] \triangleq \sum_n x[n]x[n-\ell].$$

Let X(z) be the z-transform of x[n] with ROC $\alpha < |z| < \beta$.

(a) Show that the z-transform of $r_{xx}[\ell]$ is given by $R_{xx}(z) = X(z)X(z^{-1})$.

(b) Let $x[n] = a^n u[n]$, |a| < 1. Determine $R_{xx}(z)$ and sketch its pole-zero plot and the ROC.

Solution

(a) Show that the z-transform of $r_{xx}[\ell]$ is given by $R_{xx}(z) = X(z)X(z^{-l})$. $r_{xx}[\ell] \triangleq \sum_{n} x[n]x[n-\ell] = \sum_{m} x[m+l]x[m] = x[l] * x[-l]$

By applying the folding property, the z-transform of sequence x[-l] is $X(z^{-1})$ with ROC : $\beta - 1 < |z| < \alpha - 1$. Hence, we prove $R_{xx}(z) = X(z)X(z^{-1})$ ROC : $\max\{\alpha, \beta - 1\} < |z| < \min\{\beta, \alpha - 1\}$

(b) Let $x[n] = a^n u[n]$, |a| < 1. Determine $R_{xx}(z)$ and sketch its pole-zero plot and the ROC.

$$X(z) = \frac{1}{1 - \alpha z^{-1}}, \text{ ROC} : |z| > |\alpha|$$

$$X(z^{-1}) = \frac{-\alpha z^{-1}}{1 - \alpha^{-1} z^{-1}}, \text{ ROC} : |z| < |\alpha^{-1}|$$

$$R_{zz}(z) = X(z)X(z^{-1}) = \frac{-\alpha z^{-1}}{1 - (\alpha + \alpha^{-1})z^{-1} + z^{-2}}, \text{ ROC} : |\alpha| < |z| < |\alpha^{-1}|$$

7. (12%) Determine the DTFT of following signals:

(a)
$$x1[n] = (\frac{1}{4})^n cos(\frac{\pi n}{4})u[n-2]$$

(b) $x2[n] = sin(0.1\pi n)(u[n] - u[n-10])$

Solution

DTFT :
$$X(e^{jw}) = \sum_{n=-\infty}^{n=\infty} x[n]e^{-jwn}$$

(a) $x1[n] = (\frac{1}{4})^n \cos(\frac{\pi n}{4})u[n-2]$
 $X1(e^{jw}) = \sum_{n=2}^{n=\infty} (\frac{1}{4})^n \cos(\frac{\pi n}{4})e^{-jwn}$
 $= \sum_{n=2}^{n=\infty} (\frac{1}{4})^n (\frac{1}{2})(e^{j\frac{\pi n}{4}} + e^{j\frac{-\pi n}{4}})e^{-jwn}$
 $= \sum_{n=2}^{n=\infty} (\frac{1}{4})^n (\frac{1}{2})(e^{-j(w-\frac{\pi}{4})n} + e^{-j(w+\frac{\pi}{4})n})$
 $= (\frac{1}{4})^2 (\frac{1}{2})(e^{-j(w-\frac{\pi}{4})^2} + e^{-j(w+\frac{\pi}{4})^2}) + (\frac{1}{4})^3 (\frac{1}{2})(e^{-j(w-\frac{\pi}{4})^3} + e^{-j(w+\frac{\pi}{4})^3}) + ...$
 $= (\frac{1}{2})^5 (\frac{e^{-j(w-\frac{\pi}{4})^2}}{1-\frac{1}{4}e^{-j(w+\frac{\pi}{4})^2}})$

(b)
$$x2[n] = sin(0.1\pi n)(u[n] - u[n - 10])$$

 $X2(e^{jw}) = \sum_{n=0}^{n=9} sin(0.1\pi n) e^{-jwn}$
 $= \sum_{n=0}^{n=9} (\frac{1}{2j})(e^{j0.1\pi n} - e^{-j0.1\pi n})e^{-jwn}$
 $= \sum_{n=0}^{n=9} (\frac{1}{2j})(e^{j(0.1\pi - w)n} - e^{-j(0.1\pi + w)n})$
 $= (\frac{1}{2j})(\frac{1-e^{j10(0.1\pi - w)}}{1-e^{j(0.1\pi - w)}} - \frac{1-e^{j10(-0.1\pi - w)}}{1-e^{j(-0.1\pi - w)}})$

8. (15%) Let x[n] and y[n] denote complex sequences and $X(e^{j\omega})$ and $Y(e^{j\omega})$ their respective Fourier transforms

- (a) Determine, in terms of x[n] and y[n], the sequence whose Fourier transform is
- $X(e^{j\omega})Y^*(e^{j\omega})$. (See lecture slide ch4 p21. You can use * as convolution operator)
- (b) Using the result in part (a), show that

$$\sum_{n=-\infty}^{\infty} x[n]y^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) d\omega. \quad (eq.7b)$$

(eq.7b is a more general form of Parseval's theorem, as mentioned in lecture slide ch4 p17.)

(c) Using (eq.7b), determine the value of the sum

$$\sum_{-\infty}^{\infty} \frac{\sin(\pi n/5)}{2\pi n} \frac{\sin(\pi n/3)}{7\pi n}$$

(*Hint: Check what is the expression of an inverse Fourier transform of a rectangular pulse mentioned in lecture slide ch4 p23.*)

Solution

(a) Determine, in terms of x[n] and y[n], the sequence whose Fourier transform is $X(e^{j\omega})Y^*(e^{j\omega})$. The Fourier transform of y^{*}[-n] is $Y^*(e^{j\omega})$, $G(e^{j\omega}) = X(e^{j\omega})Y^*(e^{j\omega})$

$$g[n] = x[n] * y^*[-n]$$

(b) Using the result in part (a), show that

$$\sum_{n=-\infty}^{\infty} x[n]y * [n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})Y * (e^{j\omega})d\omega. \quad (eq.7b)$$

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) e^{j\omega n} d\omega = \sum_{n=-\infty}^{\infty} (x[n] * y^*[-n]) e^{-j\omega n}$$

$$=\sum_{n=-\infty}^{\infty}\sum_{k=-\infty}^{\infty}x[k]y^{*}[k-n]e^{-j\omega n}$$

for n = 0,
$$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) d\omega = \sum_{k=-\infty}^{\infty} x[k] y^*[k]$$

(c) Using (eq.7b), determine the value of the sum

