

2013 AIC Final Exam Solution

1.

$$Z_1 = Z/(1+A)$$

$$Z_2 = Z/(1+A^{-1})$$

2.

(a)

$$C_{in} = C_{GD}(1+g_m R_D) + C_{GS} = 1010 fF$$

$$C_{out} = C_{GD}(1+(g_m R_D)^{-1}) + C_{DB} = 10.025 fF$$

(b)

$$W_{in} = 1/(R_{in} * C_{in}) = 9.901 M \text{ rad/s}$$

$$W_{out} = 1/(R_{out} * C_{out}) = 997.51 M \text{ rad/s}$$

(c)

$$W_z = g_m / C_{GD} = 400 G \text{ rad/s}$$

(d)

$R_S = 100 k\Omega$ as freq to infinity

3.

(a)

$$W_{p1} = 1/(R_{out} * C_L) = 200 M \text{ rad/s}$$

(b)

$$W_{p2} = g_m / C_E = 100 G \text{ rad/s}$$

(c)

$$W_z = 2g_m / C_E = 200 G \text{ rad/s}$$

(d)

$$W_z = 2 W_{p2}$$

4.

(a)

$$\text{Output referred noise} = 4kT(g_{m1} + g_{m2}) * (2/3)(r_{o1} || r_{o2})^2$$

(b)

$$\text{Input referred noise} = 4kT * (2/3) * (g_{m1} + g_{m2}) / (g_{m1})^2$$

(c)

Increase g_{m1} / decrease g_{m2}

5.

(a)

$$G_m R_F$$

(b)

$$G_m / (1 + G_m R_F)$$

(c)

$$R_{in} (1 + G_m R_F)$$

(d)

$$R_{out} (1 + G_m R_F)$$

(e)

Just illustrate the trade-off on G-BW between open-loop & closed-loop.

6.

視情況給分，有提到邊就給一分。

7.

$$R_{out} = A g_{m2} r_{o2} R_S = (g_{m3} r_{o3}) g_{m2} r_{o2} R_S = 2G\Omega$$

8.

(a) $\beta = -1/R_F = -20\mu(\Omega^{-1})$

(b) $A_{open} = (R_S || R_F) * -g_m (R_D || R_F) = -2.22M \text{ (V/A)}$

(c) $A_{closed} = A_{open} / 1 + \beta A_{open} = -0.489 \text{ (V/V)}$

9.

- (a)正確電路架構 sense output 電壓
- (b)negative feedback

10.

- (a) $SR = I_{SS}/C_L = 5M \text{ (V/S)}$
- (b) $SR = I_{SS}/(C_1||C_2) = 75M \text{ (V/S)}$
- (c) 參考講義 9-45
- (d) 參考講義 9-46

11.

(a)

$$W_{1st} = (R_{out1} * C_{out1})^{-1} = 200M \text{ (rad/s)}$$

$$W_{2nd} = (R_{out2} * C_{out2})^{-1} = 1G \text{ (rad/s)}$$

(b)

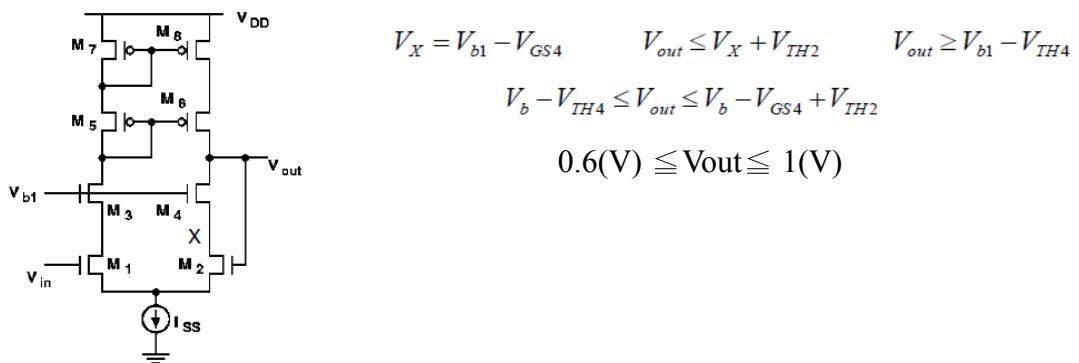
$$C_{out1'} \rightarrow 1000x$$

$$C_{out1'} = (1 + A_{V2})C_c + C_{out1}$$

$$C_c = 494.55 \text{ fF}$$

12.

(a)

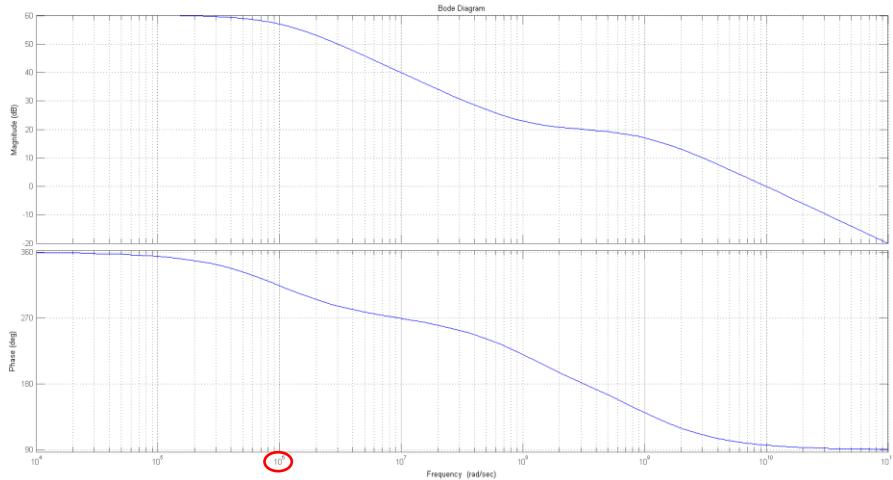


(b)

$$\text{Output swing} = V_{outmax} - V_{outmin} = 0.4(V)$$

13.

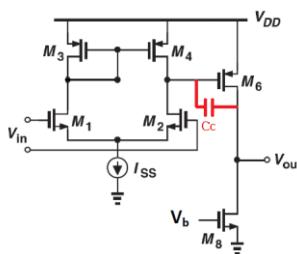
(a) pole zero 的 phase 掉 45 度, gain 分別是 +20dB/dec 及 -20dB/dec



(b) 10GHz

14.

(a)



Pole-splitting effect: Create a large capacitance, $(1+A)C_c$, at the input of the second stage amplifier to produce a dominant pole at low frequency. And increase the magnitude of the output pole by roughly a factor of $g_{m6}R_{out}$

(b)

RHP zero effect:

- 1) negative phase shift
- 2) slow down the drop of magnitude

solution:

- 1) Add a series resistor to eliminate the RHP zero or move the zero to higher frequency
- 2) Add a current buffer to break the feedthrough path

15.

(a)

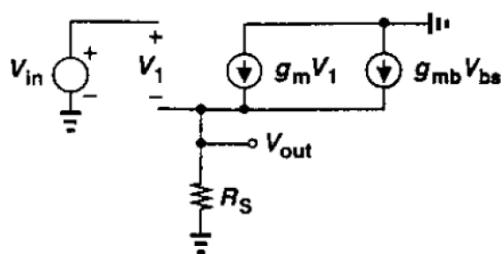
8

(b)

1.13V

16.

(a)



(b)

$$75/91 = 0.8241$$

17.

(a)

$$V_b = 1V$$

$$V_{out} = 0.4V$$

(b)

$$0.25\mu m / 2\mu m$$

18.

(a) T

(b) T

(c) F, power is double

(d) F, width only

(e) F, inversely proportional to V_{ov} because of $2I/V_{ov}$

(f) F, mobility is lower such that g_m is lower

(g) F, equal

(h) T

(i) F, except voltage swing

(j) F, impact ionization