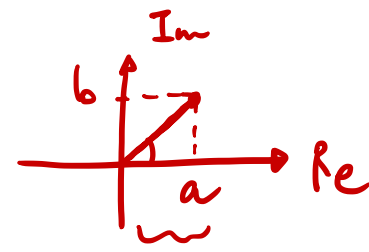




Complex Numbers

phasor



$$x = a + jb = |x| \angle x = \sqrt{a^2 + b^2} \cdot \angle \tan^{-1} \frac{b}{a}$$

$$= \underbrace{|x| \cdot \cos \angle x}_a + j \cdot \underbrace{|x| \cdot \sin \angle x}_b$$

x, y : complex numbers

$$A = x \cdot y \quad \Rightarrow \quad |A| = |x| \cdot |y|, \quad \angle A = \angle x + \angle y$$

$$B = \frac{x}{y} \quad \Rightarrow \quad |B| = \frac{|x|}{|y|}, \quad \angle B = \angle x - \angle y$$

phasor representation: represent sine wave w. respect to its mag & phase

$$v = A \cdot \cos(\omega t + \theta) \longrightarrow \begin{matrix} \text{polar} \\ \text{form} \end{matrix} A \angle \theta = \begin{matrix} \text{rectangular} \\ \text{form} \end{matrix} a + jb = \begin{matrix} \text{exponential} \\ \text{form} \end{matrix} A \cdot e^{j\theta}$$

$$1 = 1 \cdot \angle 0^\circ, \quad j = 1 \cdot \angle 90^\circ, \quad -1 = 1 \cdot \angle 180^\circ, \quad -j = 1 \cdot \angle 270^\circ = 1 \cdot \angle -90^\circ$$



o Example.

$$x(t) = 120 \cdot \sin\left(400t - \frac{\pi}{4}\right) \quad , \text{ phasor} = ?$$

$$x(t) = 120 \cdot \cos\left(400t - \frac{\pi}{4} - \frac{\pi}{2}\right) = 120 \cos\left(400t - \frac{3}{4}\pi\right)$$

$$= 120 \cdot \angle \frac{-3}{4}\pi$$

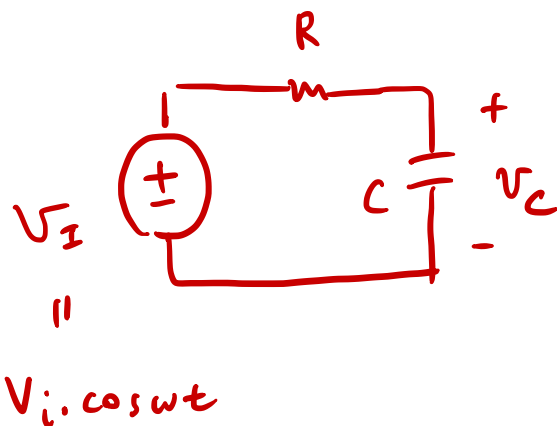
Practice : rectangular form = $a + jb$ = ?

exponential form = ?



Sinusoidal Response of RC Network

- Find $v_C(t)$ for $t \geq 0$. Assume $v_I(t) = V_i \cos(\omega t)$ for $t \geq 0$ and $v_I(t) = 0$ for $t < 0$. Initial condition $v_C(0) = 0$.



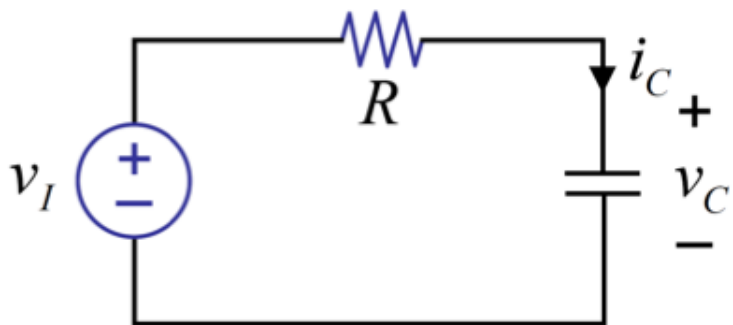


Our Usual Approach

1. Write DE for circuit by applying node method.
2. Find particular solution v_P by guessing and trial & error.
3. Find homogeneous solution v_H .
4. Total solution is $v_P + v_H$, then solve for remaining constants using initial conditions .



Step 1: Set Up the Differential Equation



$$\frac{V_I - V_C}{R} = C \cdot \frac{dV_C}{dt}$$

$$\Rightarrow RC \frac{dV_C}{dt} + V_C = V_I = V_i \cdot \cos \omega t$$



Step 2: Find the Particular Solution

$$RC \frac{dv_P}{dt} + v_P = \underline{V_i \cos(\omega t)}$$

Assume $v_P = A \cos \omega t + B \sin \omega t$



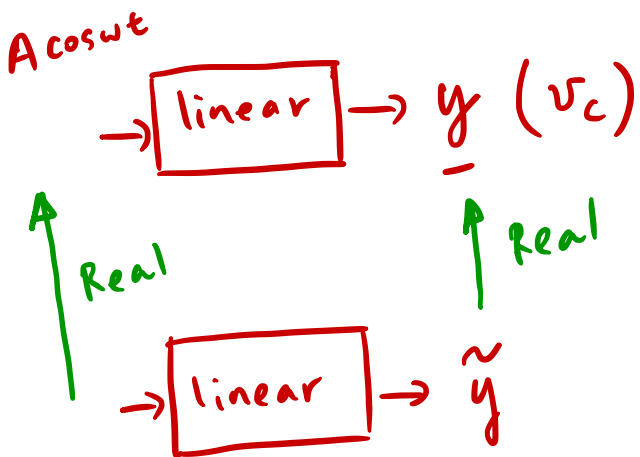


Sneaky Approach

$$RC \frac{dv_P}{dt} + v_P = V_i \cos(\omega t)$$

□ Instead of input $v_I(t) = V_i \cos \omega t$

→ Find particular solution for another input $v_{IS}(t) = V_i \exp(st)$



$$s = j\omega$$

$$RC \cdot \frac{d\tilde{v}_P}{dt} + \tilde{v}_P = V_i \cdot e^{st}$$

$$\text{Assume } \tilde{v}_P = V_P \cdot e^{st}$$

$$RC \cdot s \cdot V_P e^{st} + V_P e^{st} = V_i \cdot e^{st}$$

$$\Rightarrow e^{st} (sRC V_P + V_P) = V_i \cdot e^{st}$$

$$\Rightarrow V_P = \frac{V_i}{1 + sRC}, \quad s \neq \frac{-1}{RC}$$

$A e^{j\omega t}$

||

$A \cos \omega t + j A \sin \omega t$



Sneaky Approach

- Assume $s = j\omega$, then particular solution for input $v_{IS}(t) = V_i \exp(st)$

$$\tilde{v}_p = \frac{V_i}{1 + sRC} \cdot e^{st}$$

- Finding the particular solution to $v_{IS}(t) = V_i \exp(j\omega t)$ was easy.
- From Euler relation $v_{IS}(t) = V_i \exp(j\omega t) = V_i \cos(\omega t) + j \sin(\omega t)$

- An inverse superposition argument, assuming system is real and linear.

Sneaky Approach

$v_p(t)$ particular response to $V_i \cos \omega t$

$v_{PS}(t)$ particular response to $V_i \exp(j\omega t)$



$$v_p = \operatorname{Re}(v_p^{\sim}) = \operatorname{Re}\left(\frac{V_i}{1 + sRC} e^{st}\right) = \operatorname{Re}\left(\frac{V_i}{1 + j\omega RC} e^{j\omega t}\right)$$

$$= \operatorname{Re}\left(\frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cdot e^{j \tan^{-1}(-\omega RC)} \cdot e^{j\omega t}\right)$$

$$= \frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cos\left(\omega t + \tan^{-1}(-\omega RC)\right)$$



Step 3: Homogeneous Solution

$$RC \frac{dV_C}{dt} + V_C = 0$$

- Recalled from Chapter 10, the homogeneous solution for RC circuit v_H :

$$V_{C,H} = Ae^{st}$$

$$V_C(0) = 0$$

- The total solution:

$$v_C(t) = v_P(t) + v_H(t) = \underbrace{\frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cdot \cos(\omega t + \tan^{-1}(\omega RC))}_{v_P} + A \cdot \exp\left(\frac{-t}{RC}\right)_{v_H}$$

$$A = - \frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cdot \cos\left(-\tan^{-1}(\omega RC)\right)$$



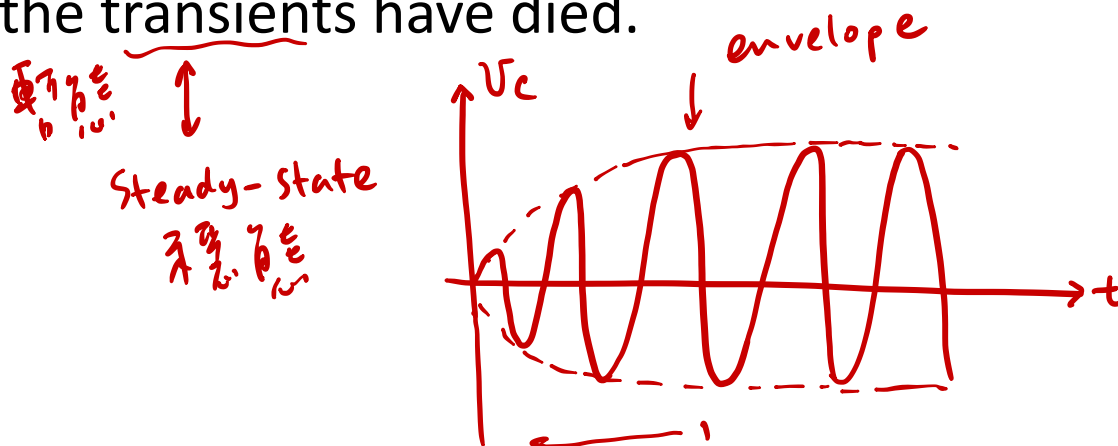
Sinusoidal Steady State

- The total solution

$$v_C(t) = \frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cos(\omega t + \phi) - \frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cos \phi \cdot e^{-\frac{t}{RC}}$$

where $\phi = \tan^{-1}(-\omega RC)$

- We are usually interested only in the particular solution for sinusoids, i.e., after the transients have died.



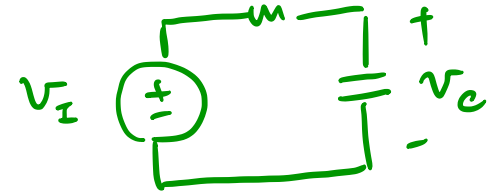
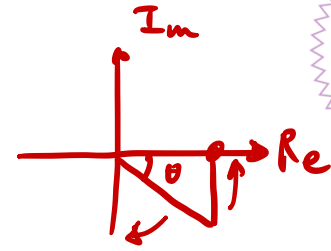
- When $t \rightarrow \infty$,

$$v_C(t) = \frac{V_i}{\sqrt{1 + (\omega RC)^2}} \cdot \cos(\omega t + \tan^{-1}(\omega RC))$$

Sinusoidal Steady State



$$V_c = \frac{V_i}{\sqrt{1+(wRC)^2}} \cdot \cos(\omega t + \theta), \quad \theta = \tan^{-1}(-wRC)$$



when

$$1) \quad \omega \rightarrow 0, \quad \theta = \tan^{-1}(-wRC) = 0$$

$$V_c = \frac{V_i}{\sqrt{1+(wRC)^2}} \cos(\omega t + \theta) = \frac{V_i}{1} \cos(\omega t) = V_i$$

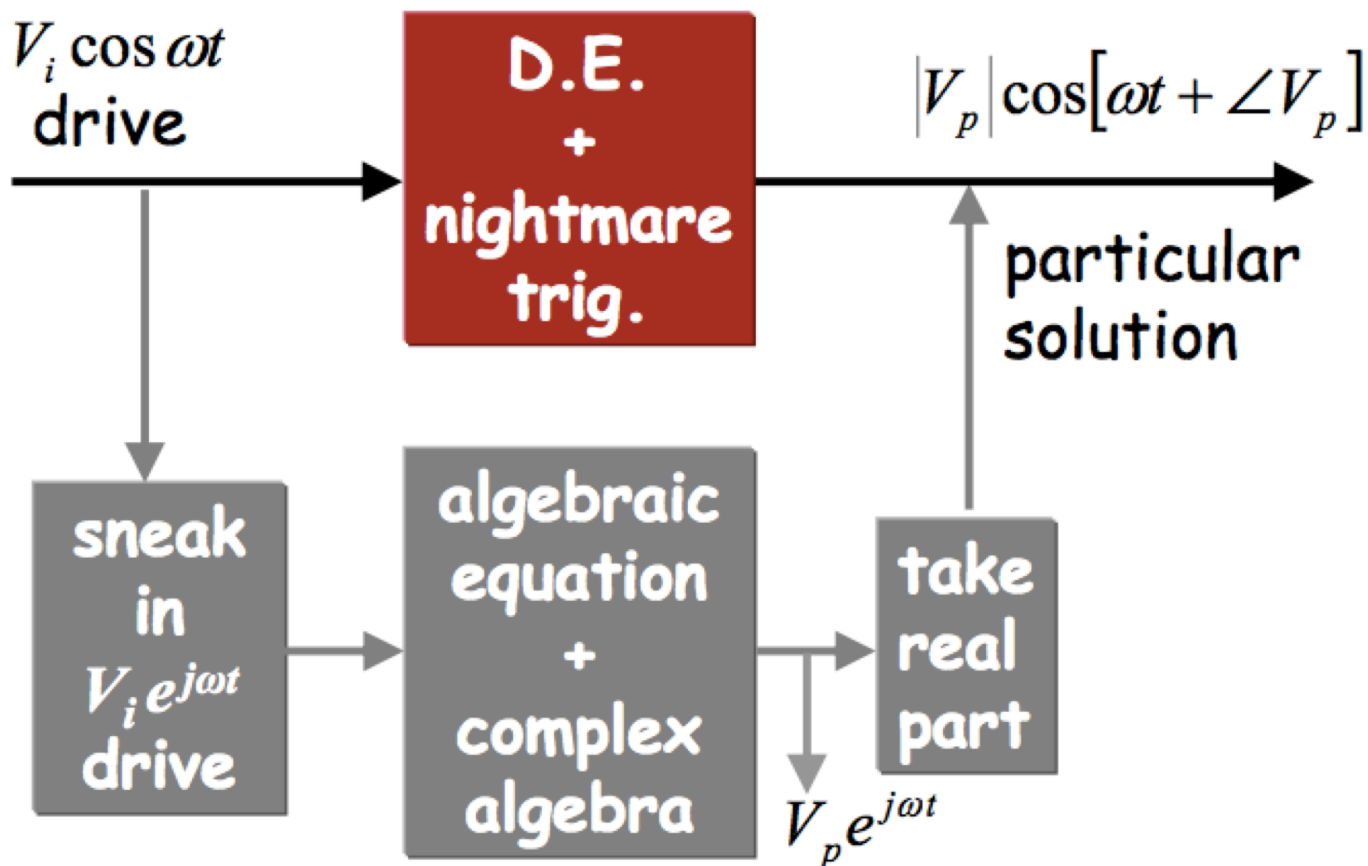
$$2) \quad \text{when } \omega \rightarrow \infty, \quad V_c = \frac{V_i}{\sqrt{1+(wRC)^2}} \cos(\omega t + \theta) = \frac{V_i}{wRC} \cos(\omega t - 90^\circ) \approx 0$$

Low-pass



Visualizing Sinusoidal Steady State

- The process of finding the particular solution v_P .





Summary of SSS Approach

1. Set up differential equation. (KCL, KVL, node method, ...)
2. Apply sneaky input $V_i \cdot e^{j\omega t}$ then find particular solution

$$RC \frac{dv_P}{dt} + v_P = V_i \cos(\omega t)$$

Assume

$$v_{PS}(t) = V_P \cdot \exp(j\omega t)$$

=

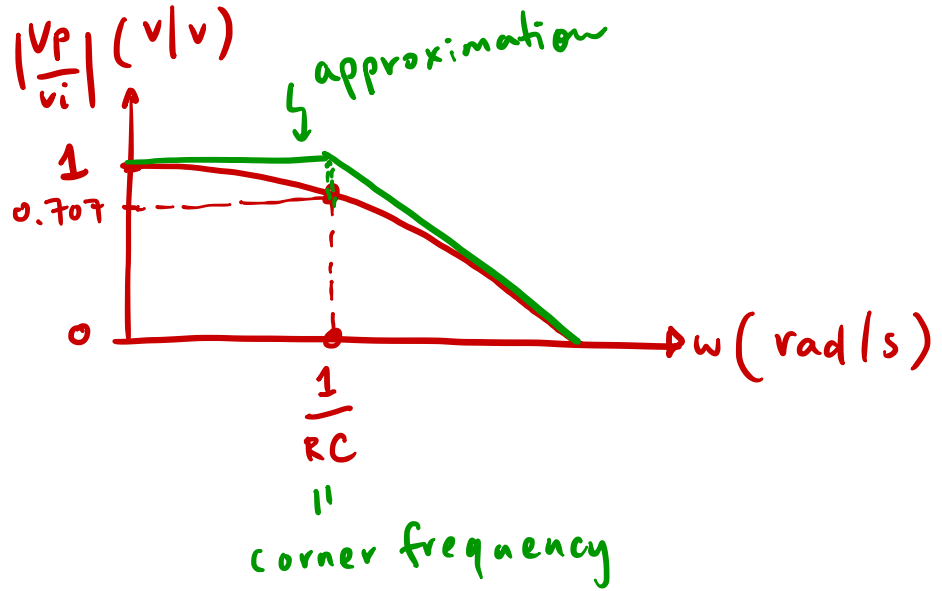


Magnitude Plot

$$= \frac{v_i / \sqrt{1 + (\omega RC)^2}}{v_i} = H$$

- Transfer function $\frac{V_P}{V_i} = \frac{1}{1 + j\omega RC} = \frac{1}{\sqrt{1 + (\omega RC)^2}} \cdot \exp(j \cdot \tan^{-1}(-\omega RC))$
 - Also known as a system function, is the ratio of the complex amplitude of the network output to the complex amplitude of the input.

Magnitude plot: *low-pass*



- At low freq, $(\omega RC)^2 \ll 1$, $|H| \sim 1$
- At high freq, $(\omega RC)^2 \gg 1$, $|H| \sim \frac{1}{\omega RC} \rightarrow 0$
- At $\omega = \frac{1}{RC}$, $|H| = \frac{1}{\sqrt{1+1}} = \frac{1}{\sqrt{2}} = 0.707$

Output voltage falls off at high frequencies!!

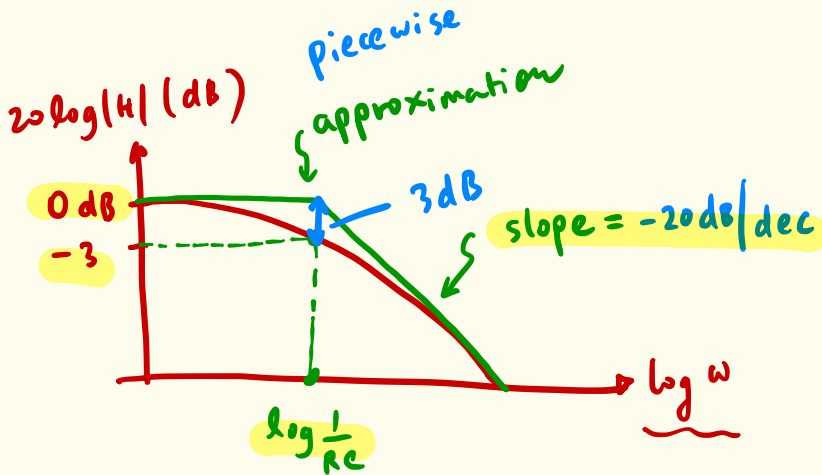
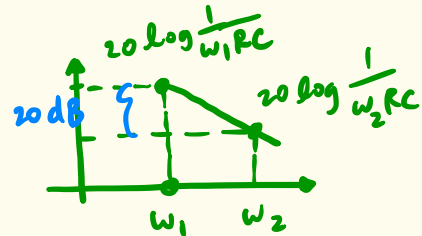
• Decibel : $\left| \frac{V_p}{V_i} \right| (v/v) \rightarrow 20 \log \left| \frac{V_p}{V_i} \right| \text{ (dB)}$

$$\begin{aligned} \log w_1 \\ \log(10w_1) \\ = \log 10 + \log w_1 \\ = 1 + \log w_1 \end{aligned}$$

• Low freq, $|H| \rightarrow 1$ $20 \log 1 = \underline{0 \text{ dB}}$

• High freq, $|H| \sim \frac{1}{\omega RC}$ $20 \log \frac{1}{\omega RC} = -20 \log \omega RC$

• Corner freq, $|H| = \frac{1}{\sqrt{2}}$ $20 \log \frac{1}{\sqrt{2}} = -3 \text{ dB}$



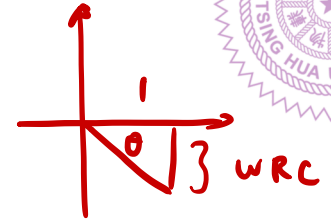
Assume $w_2 = 10w_1$ $10 \times$ (decade)

$$\begin{aligned} 20 \log \frac{1}{w_2 RC} &= 20 \log \frac{1}{10w_1 RC} \\ &= 20 \log \frac{1}{10} + 20 \log \frac{1}{w_1 RC} \\ &= -20 + 20 \log \frac{1}{w_1 RC} \end{aligned}$$



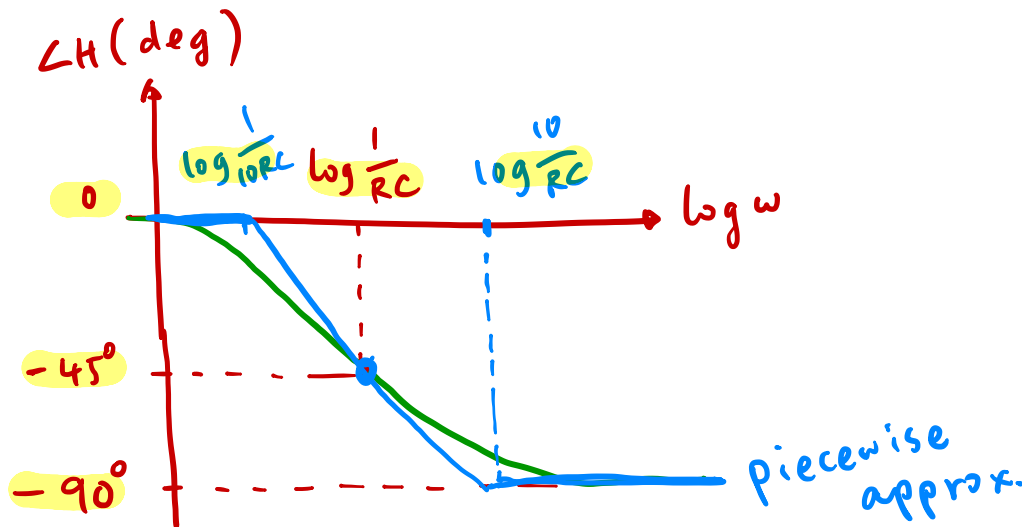
Phase Plot

□ Transfer function $H(j\omega) = \frac{V_p}{V_i} = \frac{1}{1 + j\omega RC}$



$$\angle H = \tan^{-1}(-\omega RC)$$

- At low freq, $\angle H \sim 0^\circ$
- At high freq, $\angle H \sim -90^\circ$
- At $\omega = \frac{1}{RC}$, $\angle H = -45^\circ$



Example 1.

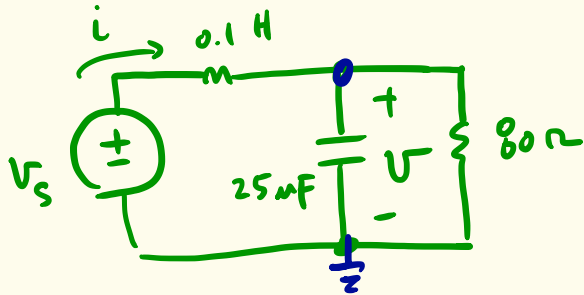
Practice.

$$V_1 = -1.796 + j3.852$$

Example 2.

$$v_s = 48 \cdot \cos(500t + 75^\circ)$$

Find $v(t)$ and $i(t)$, SSS



$$i = 25 \mu \cdot \frac{dv}{dt} + \frac{v}{80} \quad (1)$$

$$v_s - v = v_L = 0.1 \frac{di}{dt} \quad (2)$$

Use (1) in (2) $\Rightarrow v_s - v = 0.1 \left(25 \mu \frac{d^2 v}{dt^2} + \frac{1}{80} \frac{dv}{dt} \right)$

$$\Rightarrow v_s = 25 \times 10^{-7} \frac{d^2 v}{dt^2} + \frac{0.1}{80} \frac{dv}{dt} + v$$

Practice

$$v_H = ?$$

$$U_s \rightarrow \tilde{v}_s = 40 \cdot e^{j(500t + 75^\circ)}$$

Assume $\tilde{v} = Ae^{st}$

$$\tilde{v}_s = 25 \times 10^{-7} \frac{d^2 \tilde{v}}{dt^2} + \frac{0.1}{80} \frac{d\tilde{v}}{dt} + \tilde{v}$$

$$40 \cdot e^{j(500t + 75^\circ)} = 25 \times 10^{-7} \cdot A \cdot s^2 \cdot e^{st} + \frac{0.1}{80} A s e^{st} + A e^{st}$$

$$= A e^{j500t} \left(25 \times 10^{-7} \times \omega^2 + \frac{0.1}{80} \cdot j500 + 1 \right) \quad \begin{matrix} s = j\omega \\ = j \cdot 500 \end{matrix}$$

$$\Rightarrow A = 63.4 + j18.2$$

$$\tilde{v} = (63.4 + j18.2) e^{j500t} \xrightarrow{\text{Real}} v = \sqrt{(63.4)^2 + (18.2)^2} \cos\left(500t + \tan^{-1} \frac{18.2}{63.4}\right)$$

$$= 66 \cos(500t + 16^\circ)$$

$$i = 25\mu \frac{dv}{dt} + \frac{v}{80} \rightarrow \tilde{i} = 25\mu \cdot \frac{d\tilde{v}}{dt} + \frac{\tilde{v}}{80}$$

$$\tilde{v} = 66 e^{j16^\circ} e^{j500t}$$

$$\Rightarrow \tilde{i} = \underbrace{25\mu \cdot 66 \cdot e^{j16^\circ} \cdot j500 \cdot e^{j500t}} + \frac{1}{80} \cdot 66 e^{j16^\circ} e^{j500t}$$

$$= j0.825 e^{j16^\circ} e^{j500t} + 0.825 e^{j16^\circ} e^{j500t}$$

$$= 0.825 e^{j500t} (e^{j90^\circ} e^{j16^\circ} + e^{j16^\circ})$$

$$= 0.825 e^{j500t} (e^{j106^\circ} + e^{j16^\circ})$$

Real
→

$$i = 0.825 \cos(500t + 106^\circ) + 0.825 \cos(500t + 16^\circ)$$