
EE2210 Lecture 8A: Second-Order Systems

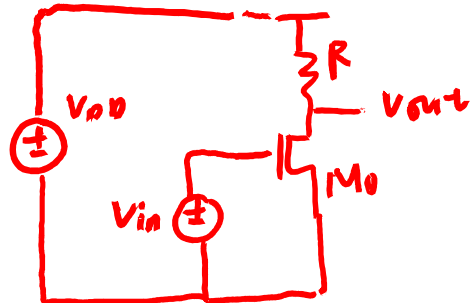
Chapter 12 and Appendix C of textbook

Ping-Hsuan Hsieh (謝秉璇)

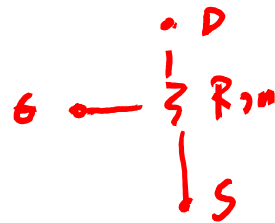
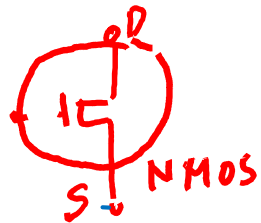
Delta Building R908

EXT 42590

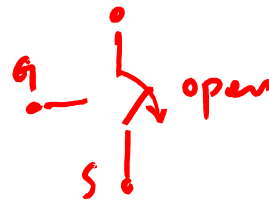
phsieh@ee.nthu.edu.tw



inverter
-DO-



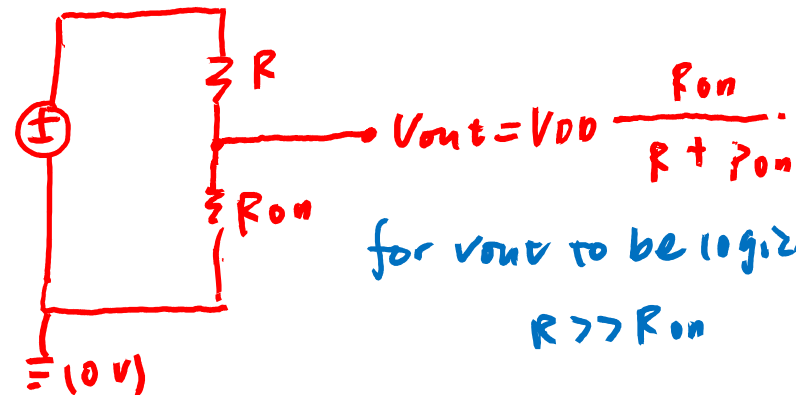
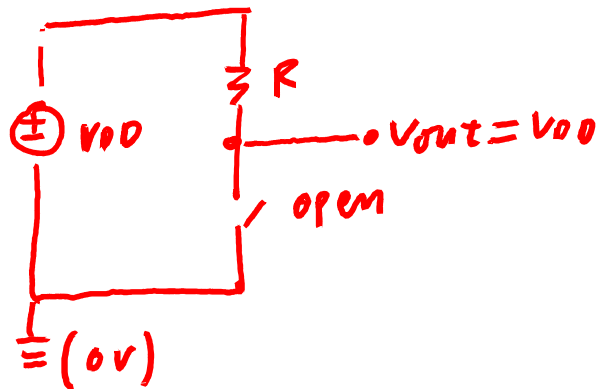
$$V_{GS} > V_{th}$$



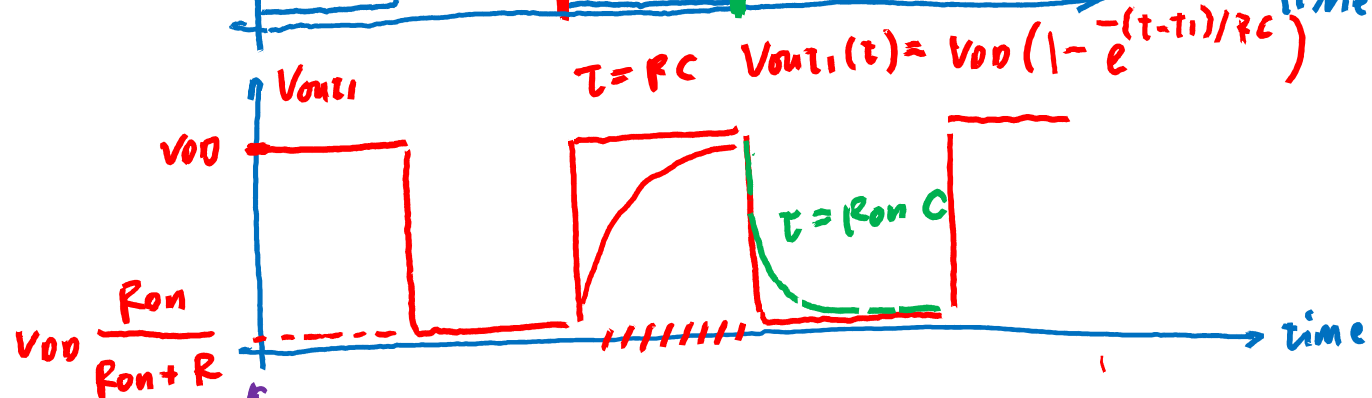
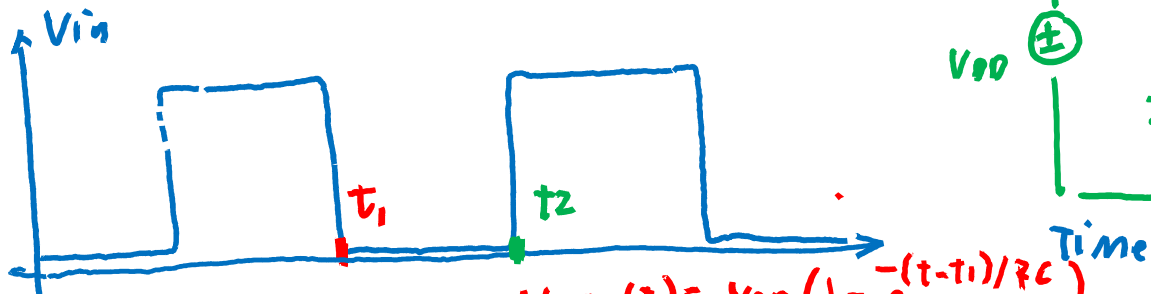
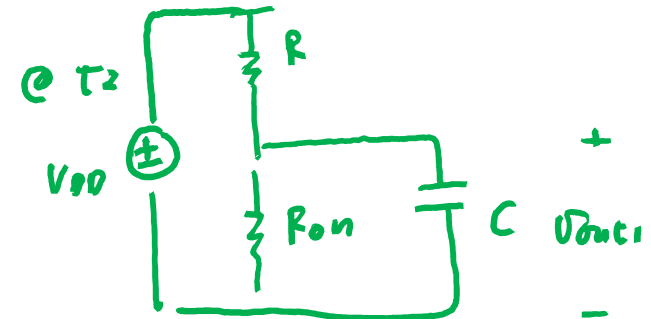
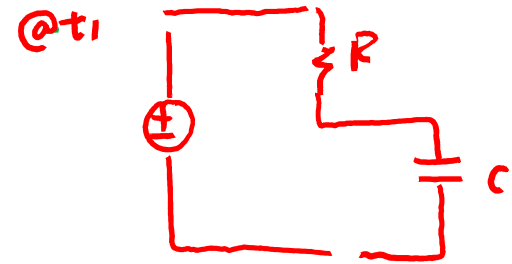
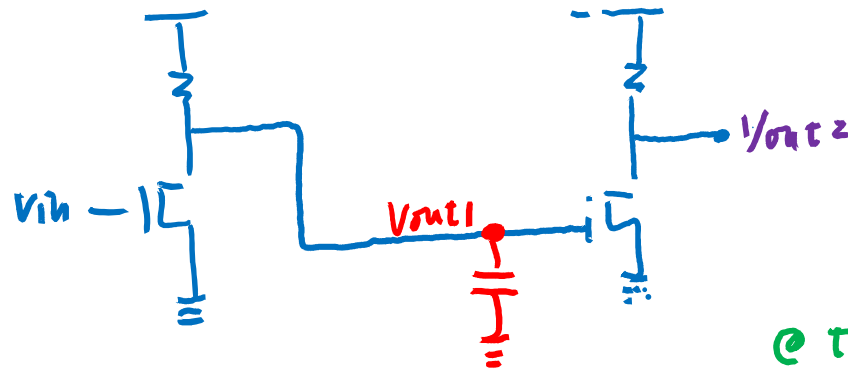
$$V_{GS} < V_{th}$$

#1 $V_{in} < V_{th}$ Mo OFF

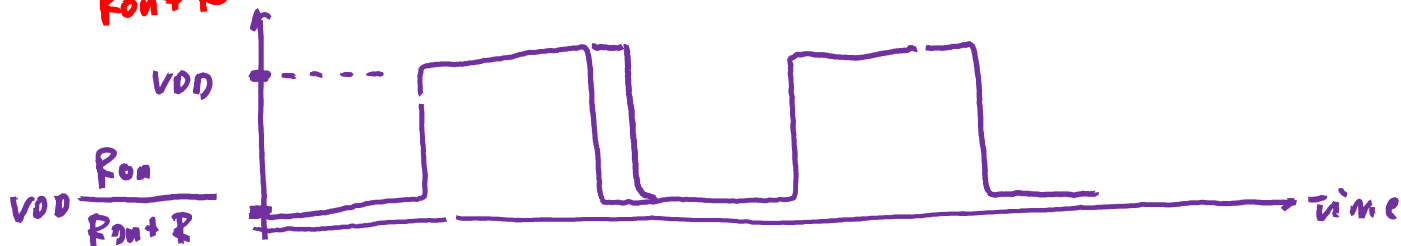
#2 $V_{in} > V_{th}$



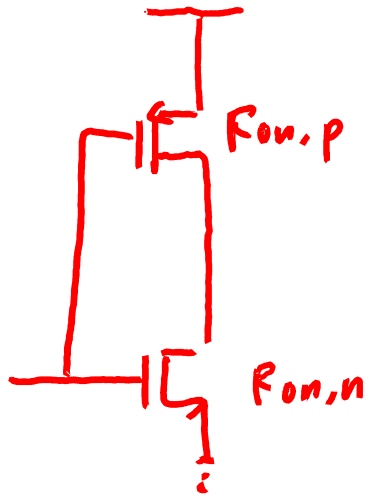
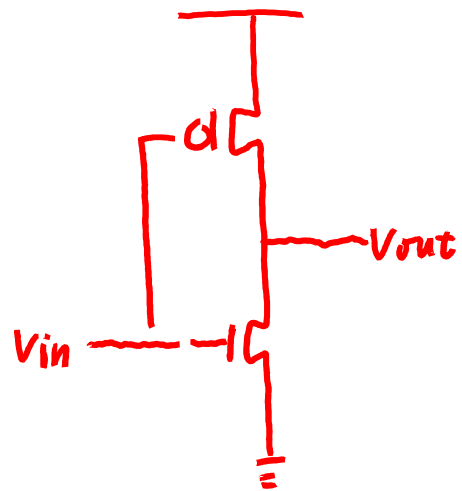
for V_{out} to be logical low
 $R \gg R_{on}$



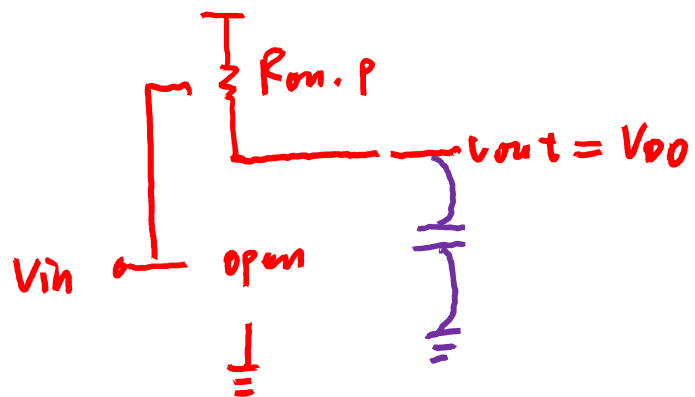
$V_{out1}(t_2^-) = V_{DD}$



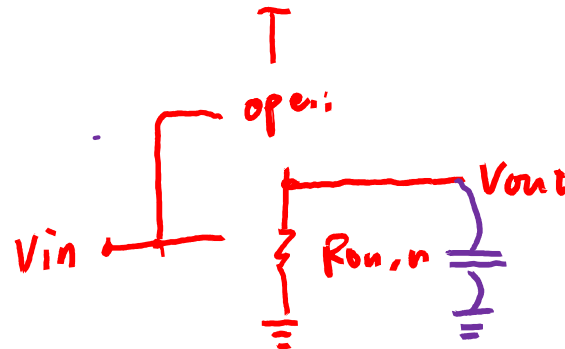
CMOS



#1 $V_{in} < V_{th,n}$

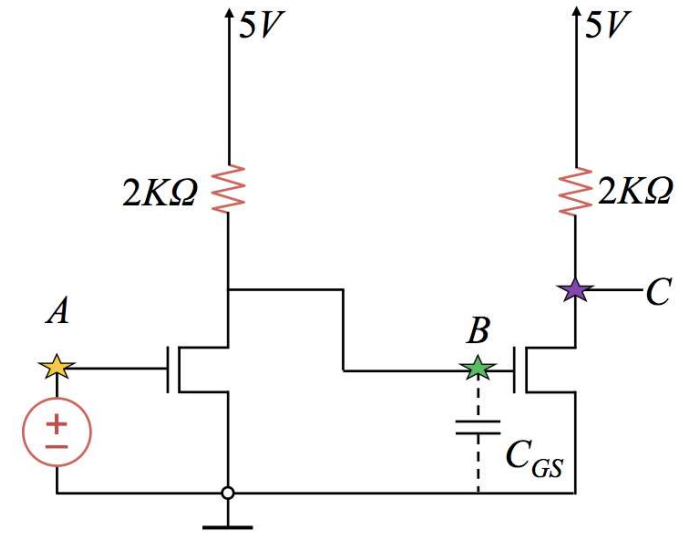
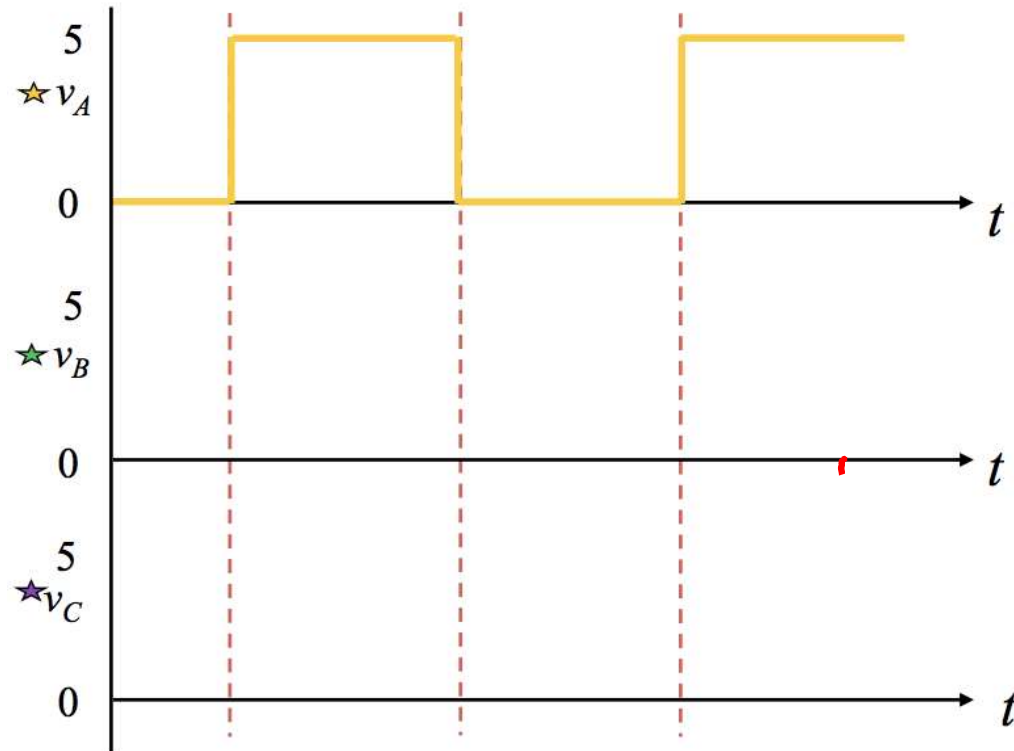


#2 $V_{in} > V_{th,n}$
 $(V_{DD} - V_{in} < |V_{th,p}|)$



An Inverter Chain Example

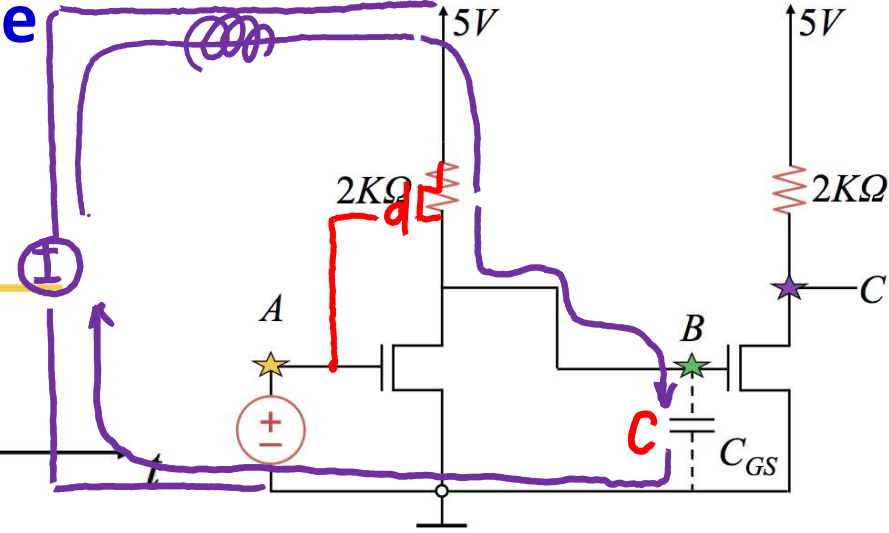
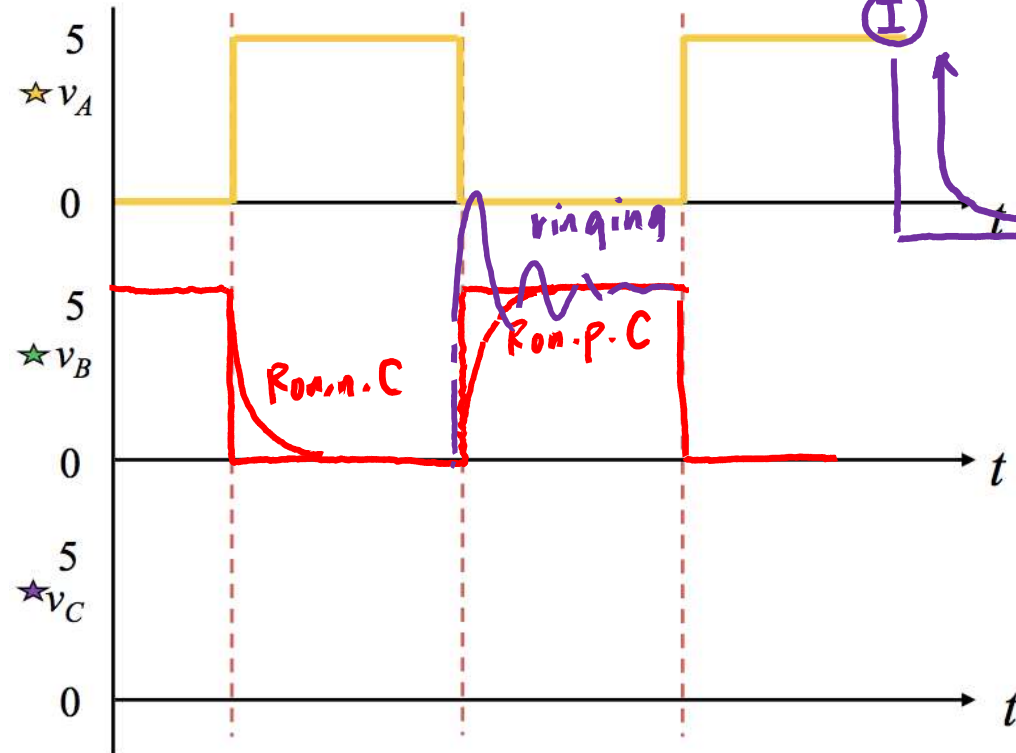
- ◆ With C_{GS} of NMOS on node B



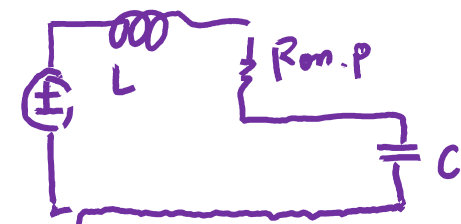
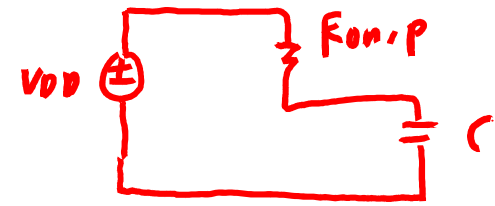
- Slow rise time and delay in signal
- We can try to increase the speed with a smaller R_L (from 2 k Ω to 50 Ω)

Observed Output – Fast Case

- ◆ With C_{GS} of NMOS on node B



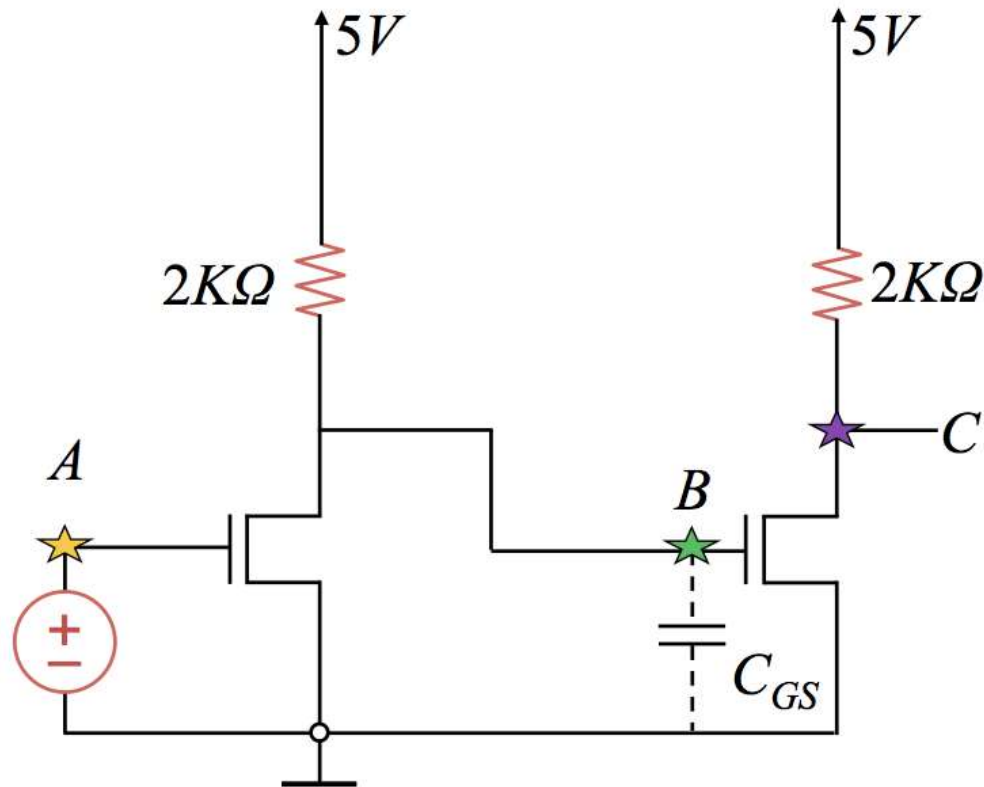
ringing
CMOS
inverter



– Ringing behavior observed!!

Fast Case – What's Really Going On

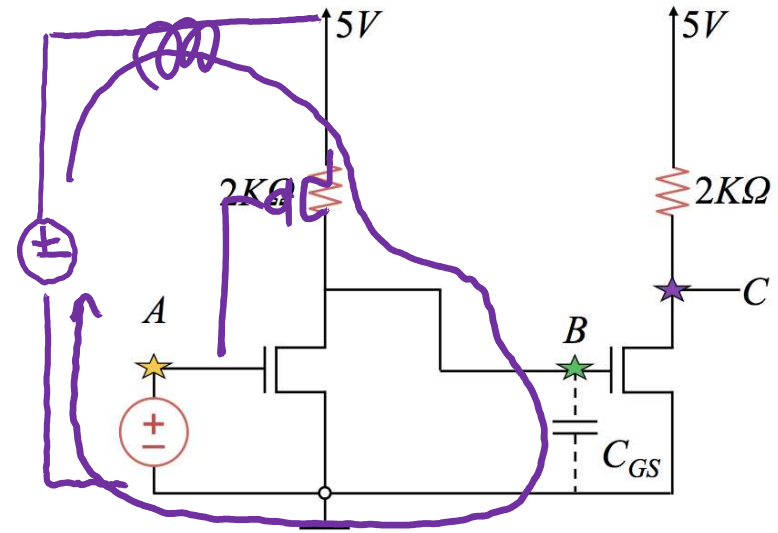
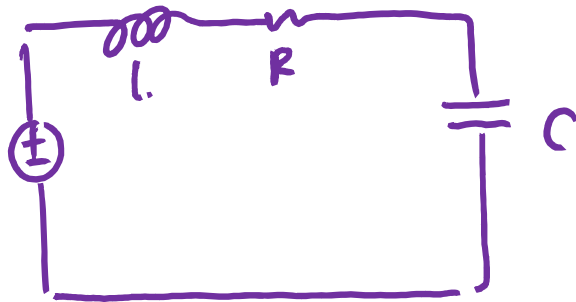
- ◆ The parasitic inductance of the wire is included



|

Second-Order Systems

- ◆ Involving R , L , C circuit elements
- ◆ Relevant circuit model

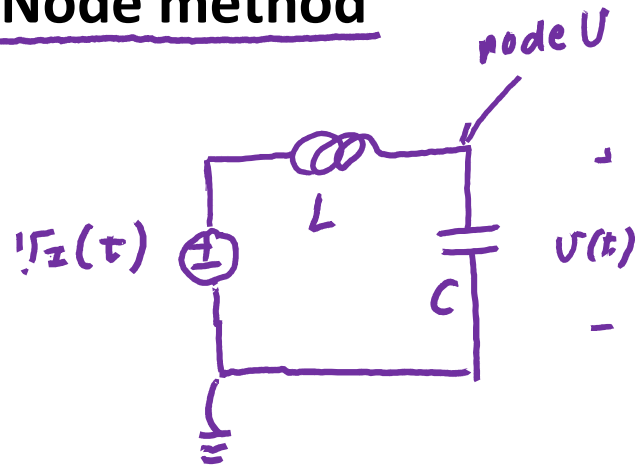


- Two energy-storage elements → second-order system

First, Let's Analyze the LC Network

◆ We will introduce R into the circuit later

◆ Node method



① reference point

② KCL @ node V

for C:

$$i_C(t) = C \frac{dv(t)}{dt}$$

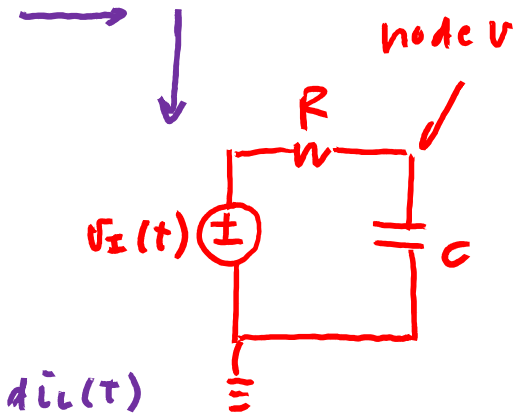
for L:

$$(v_I(t) - v(t)) = L \frac{di_L(t)}{dt}$$

$$\Rightarrow \text{KCL: } \frac{1}{L} \int_{-\infty}^t (v_I(t) - v(t)) dt = C \frac{dv(t)}{dt}$$

$$LC \frac{d^2v(t)}{dt^2} + v(t) = v_I(t)$$

$$LC: \frac{d^2}{dt^2} \text{ s}^{-2}$$



$$RC \frac{dv(t)}{dt} + v(t) = v_I(t)$$

$$RC: \frac{d}{dt} \text{ s}$$

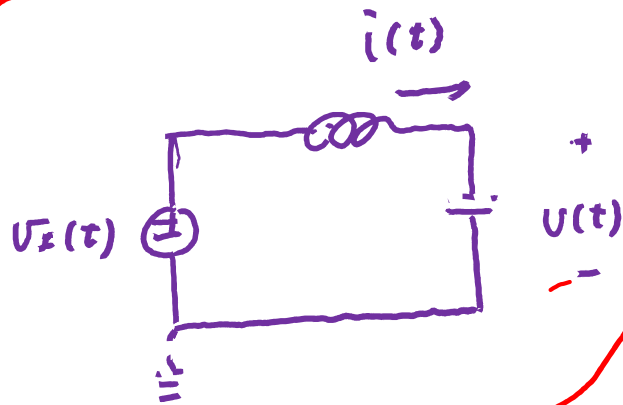
Method of Particular and Homogeneous Solutions

◆ Four-step procedure

$$LC \frac{d^2 v}{dt} + v = v_I$$

1. Find the particular solution $v_P(t)$
2. Find the homogeneous solution $v_H(t)$
 - Four-step procedure
3. The total solution is the sum of the particular solution and homogeneous solution
4. Use initial condition to solve for the remaining constraints

Let's Solve



$$L i \frac{d^2 v(t)}{dt^2} + v(t) = v_I(t)$$

$$\Rightarrow \text{求 } v(t) = f(L, C, v_I(t))$$

For step input

$$v_I(t) = V_I \cdot u(t)$$



and for initial conditions

$$\text{(zero-state)} \quad \begin{cases} v(0) = 0 \\ i(0) = 0 \end{cases}$$

求其 response $v(t)$ for $t \geq 0$

1. Particular Solution $v_p(t)$

$$(v(t) = v_p(t) + v_h(t))$$

For $t \geq 0$

$$LC \frac{d^2 v_p(t)}{dt^2} + v_p(t) = 1 \text{ V}$$

猜 $v_p(t) = V \text{ I}$

(t) [R] 去有合, okay !!

2. Homogeneous Solution

◆ Look for solution to $L \frac{d^2 v_H(t)}{dt^2} + v_H(t) = 0$

◆ Four-step method:

(2A) 猜 $v_H(t) = A e^{st}$ (A, s 待求) $A \neq 0$

(2B) 代入 $LC \cdot A s^2 e^{st} + A e^{st} = 0$

$$\Rightarrow LC \cdot s^2 + 1 = 0 \text{ (char. eq.)}$$

$$RC \cdot s + 1 = 0$$

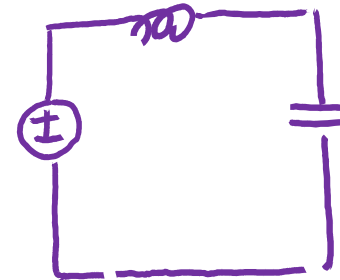
(2C) 求 s 解 $\therefore s = \pm j \sqrt{\frac{1}{LC}}$ ($j = \sqrt{-1}$, Appendix C)

$$\text{if } \sqrt{\frac{1}{LC}} = \omega_0 \Rightarrow s = \pm j\omega_0$$

(2D) general solution $v_H(t) = A_1 e^{j\omega_0 t} + A_2 e^{-j\omega_0 t}$

3. Total Solution

$$\begin{aligned}
 v(t) &= v_p(t) + v_H(t) \\
 &= V_I + A_1 e^{j\omega_0 t} + A_2 e^{-j\omega_0 t}
 \end{aligned}$$



4. Find unknowns from initial conditions

$$\left(\begin{array}{l} v(\emptyset) = \emptyset \\ \dot{v}(\emptyset) = \emptyset \end{array} \right. \begin{array}{l} \therefore \emptyset = V_I + A_1 + A_2 \\ i(t) = C \frac{dv(t)}{dt} \\ = C \cdot A_1 (j\omega_0) - C A_2 (j\omega_0) = \emptyset \end{array} \left. \vphantom{\left(\begin{array}{l} v(\emptyset) = \emptyset \\ \dot{v}(\emptyset) = \emptyset \end{array} \right.} \right) \quad A_1 = A_2 = -\frac{V_I}{2}$$

$$v(t) = V_I - \frac{V_I}{2} \left(e^{j\omega_0 t} + e^{-j\omega_0 t} \right)$$

3. Total Solution

◆ Remember Euler relation:

$$e^{jx} = \cos x + j \sin x$$

$$V_i - \frac{V_i}{2} (e^{j\omega_0 t} + e^{-j\omega_0 t})$$

$$= V_i - \frac{V_i}{2} (\cos \omega_0 t + \cancel{j \sin \omega_0 t} + \cos \omega_0 t - \cancel{j \sin \omega_0 t})$$

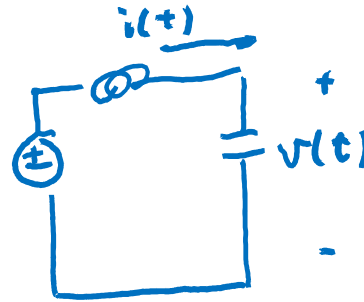
$$v(t) = \underbrace{V_i}_{\text{particular solution}} - \underbrace{V_i \cdot \cos \omega_0 t}_{\text{homogeneous solution}}$$

↑
particular
solution

homogeneous solution

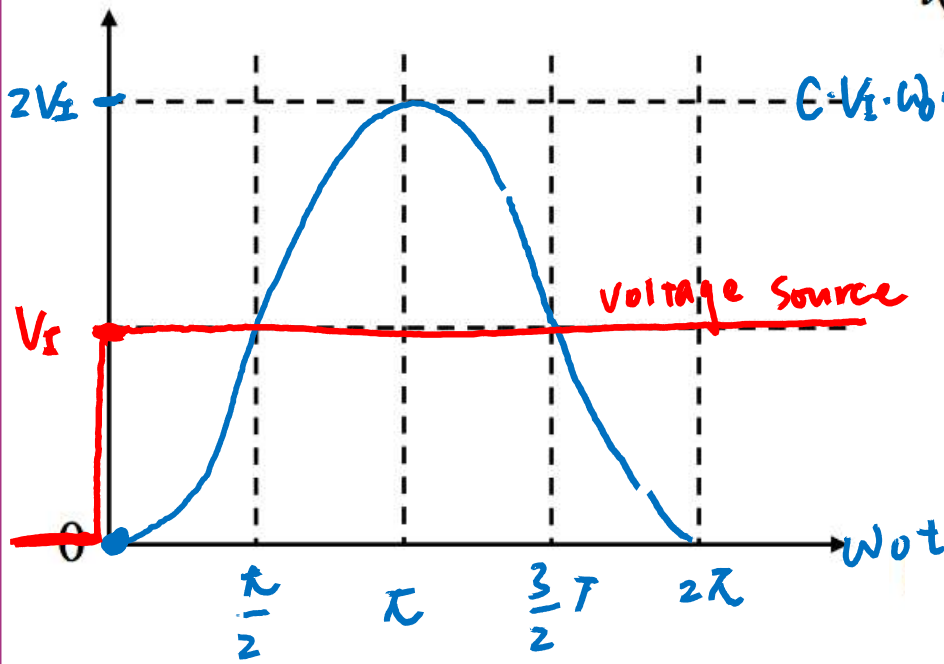
the output looks sinusoidal

$$i(t) = C \frac{dv(t)}{dt} = C \cdot V_i \cdot \omega_0 \cdot \sin(\omega_0 t)$$

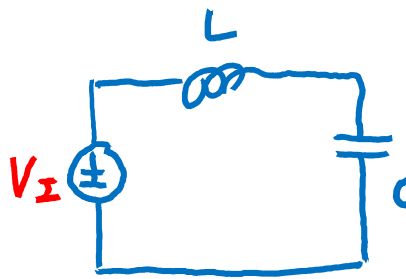
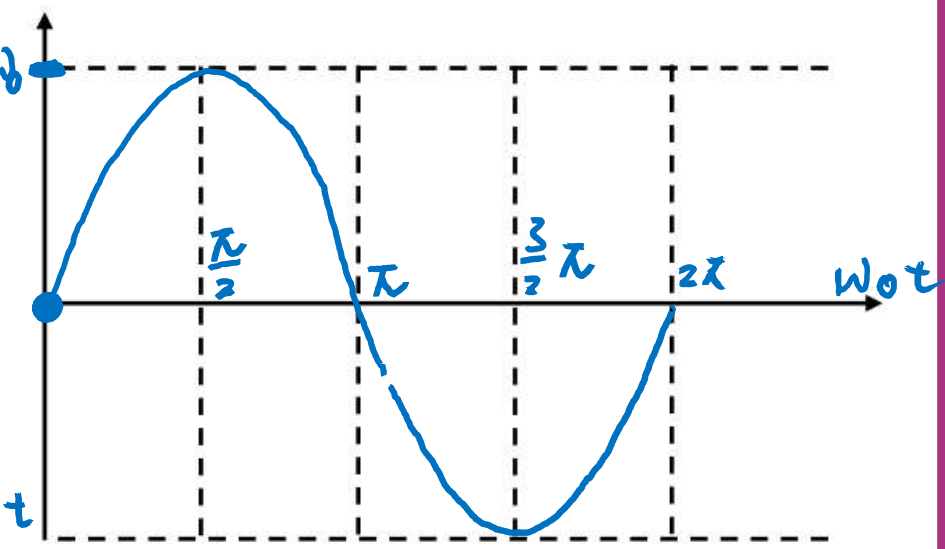


Plotting the Total Solution

$$v(t) = V_I - V_I \cdot \cos \omega_0 t$$



$$i(t) = C \cdot V_I \cdot \omega_0 \cdot \sin(\omega_0 t)$$



$$v(\phi) = \phi$$

$$i(\phi) = \phi$$

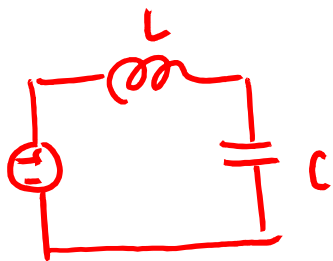
$$Q = CV$$

$$i_C(t) = C \frac{dv_C(t)}{dt}$$

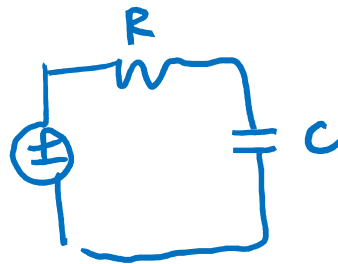
$$v_L(t) = L \frac{di_L(t)}{dt}$$

Summary of Method

1. Write DE for circuit by applying node method
2. Find particular solution v_P by guessing and trial & error
3. Find homogeneous solution v_H
 - A. Assume solution of the form Ae^{st}
 - B. Obtain characteristic equation $LCs^2 + 1 = 0$ $RCs + 1 = 0$
 - C. Solve characteristic equation for roots s_i
 - D. Find v_H by summing $A_i \exp(s_i t)$ terms
4. Total solution is $v_P + v_H$, then solve for remaining constants using initial conditions



$$LC \frac{d^2 v(t)}{dt^2} + v(t) = VI$$

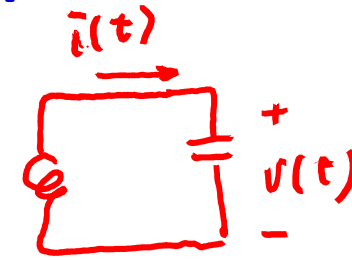


Example – Undriven LC Network Response

- ◆ What if we have: $V_I = 0$

zero-input response

- ◆ We can obtain the answer directly from the homogeneous solution (with $v_I = 0$)



$$V_I = 0$$

$$\begin{cases} v(0) = V \\ i(0) = 0 \end{cases}$$

$$LC \frac{d^2 v(t)}{dt^2} + v(t) = 0$$

$$\frac{d^2}{dt^2} A \cdot e^{st} \quad A \neq 0 \quad \text{root } \left[\frac{d^2}{dt^2} s = \pm j\sqrt{\frac{1}{LC}} = \pm j\omega_0 \right]$$

$$v(t) = A_1 \cdot e^{+j\omega_0 t} + A_2 \cdot e^{-j\omega_0 t}$$

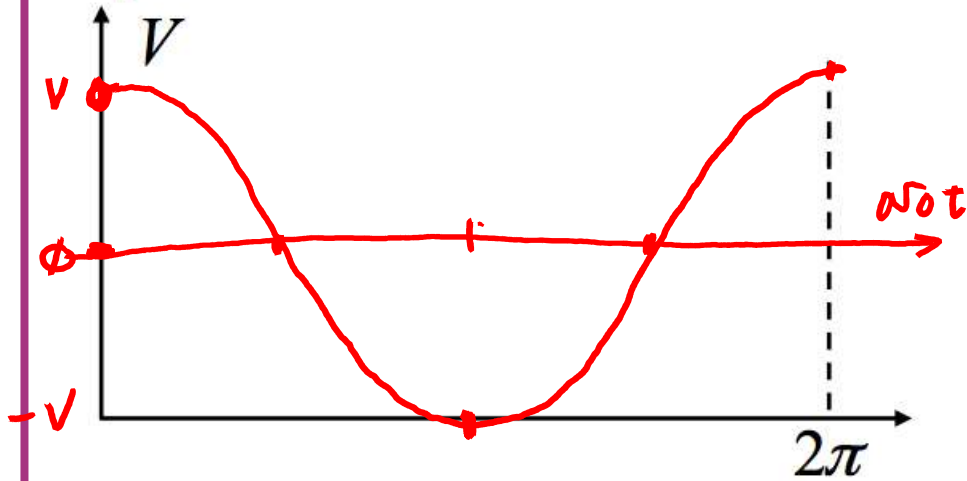
initial condition 1) $V = A_1 + A_2$

2) $i(t) = C \cdot \frac{dv(t)}{dt} = C \cdot (j\omega_0) A_1 e^{j\omega_0 t} - C(j\omega_0) A_2 e^{-j\omega_0 t}$

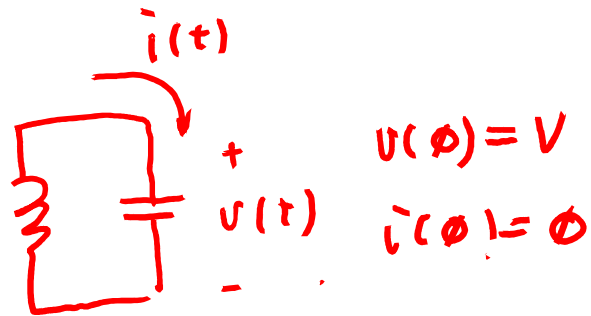
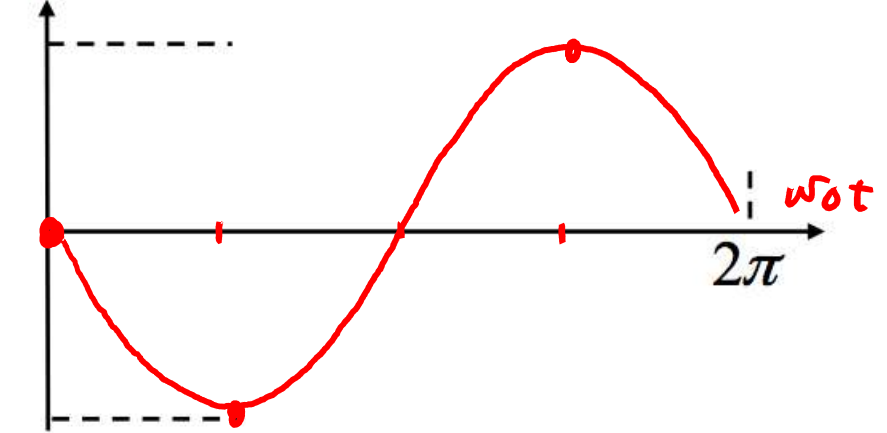
$$i(0) = 0 = A_1 - A_2 \Rightarrow v(t) = \frac{V}{2} e^{j\omega_0 t} - \frac{V}{2} e^{-j\omega_0 t} = V \cdot \cos \omega_0 t$$

Undriven LC Network Response

$$v_C = V \cdot \cos \omega_0 t$$

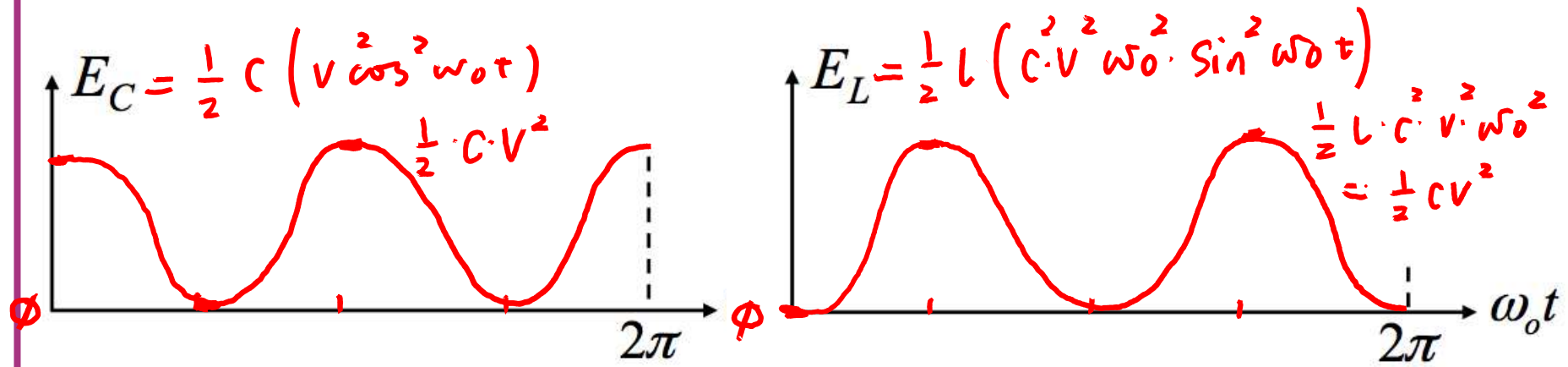


$$i_C = -C V \cdot \omega_0 \cdot \sin \omega_0 t$$



C

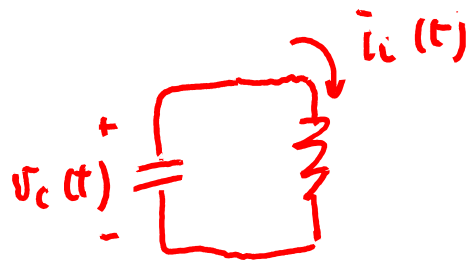
Energy



$$E_C(t) + E_L(t) = \frac{1}{2} C V^2$$

◆ Notice:

- ◆ Total energy in the system is a constant, but it goes back and forth between the capacitor and the inductor.



$$\begin{cases} -i_L(t) = C \frac{dV_c(t)}{dt} \\ V_c(t) = L \frac{di_L(t)}{dt} \end{cases}$$

$$x \neq \frac{1}{A} \quad i_L(t) = \beta \cdot \sin(\omega t)$$

$$V_c(t) = A \cos(\omega t)$$