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EE 3640 Communication Systems I Spring 2023

## Homework Assignment No. 4 Due 1:20pm, May 15, 2023

Reading: Haykin & Moher, Chapter 5.

## **Problems for Solution:**

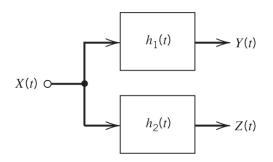
1. Let X and Y be independent Gaussian random variables, each with zero mean and unit variance. Define the random process

$$Z(t) = X\cos(2\pi t) + Y\sin(2\pi t).$$

- (a) Let  $Z_1$  and  $Z_2$  be the random variables obtained by observing Z(t) at times  $t_1$  and  $t_2$ , respectively. Find the joint probability density function of  $Z_1$  and  $Z_2$ . Assume that  $\sin[2\pi(t_1 - t_2)] \neq 0$ .
- (b) Is the process Z(t) wide-sense stationary? Why?
- 2. Consider two linear filters connected in cascade as shown below. Let X(t) be a widesense stationary process with autocorrelation function  $R_X(\tau)$ . The random process appearing at the first filter output is V(t) and that at the second filter output is Y(t).

$$X(t) \longrightarrow \begin{array}{c} h_1(t) \\ & & \\ \end{array} \begin{array}{c} V(t) \\ & & \\ \end{array} \begin{array}{c} h_2(t) \\ & & \\ \end{array} \begin{array}{c} Y(t) \\ & & \\ \end{array}$$

- (a) Find the autocorrelation function of Y(t).
- (b) Find the cross-correlation function  $R_{VY}(\tau)$  of V(t) and Y(t).
- 3. A stationary Gaussian process X(t) with mean  $\mu_X$  and variance  $\sigma_X^2$  is passed through two linear filters with impulse responses  $h_1(t)$  and  $h_2(t)$ , yielding processes Y(t) and Z(t).



(a) Find the covariance of the random variables  $Y(t_1)$  and  $Z(t_2)$ , for some times  $t_1$  and  $t_2$ . Express your answer in terms of  $h_1(t)$ ,  $h_2(t)$ , and the autocovariance function of X(t) given by

$$C_X(\tau) = \mathbb{E}\left[ (X(t+\tau) - \mu_X)(X(t) - \mu_X) \right].$$

- (b) Let  $H_1(f)$  and  $H_2(f)$  be the corresponding transfer functions of  $h_1(t)$  and  $h_2(t)$ , respectively. Show that  $Y(t_1)$  and  $Z(t_2)$  are statistically independent if  $H_1(f)$ and  $H_2(f)$  are non-overlapping, i.e.,  $H_1(f) = 0$  for all f such that  $H_2(f) \neq 0$ , and  $H_2(f) = 0$  for all f such that  $H_1(f) \neq 0$ .
- 4. A white noise w(t) of power spectral density  $N_0/2$  is applied to a *Butterworth* low-pass filter of order n, whose amplitude response is given by

$$|H(f)| = \frac{1}{\left[1 + (f/f_0)^{2n}\right]^{1/2}}.$$

- (a) Determine the noise equivalent bandwidth for this low-pass filter.
- (b) What is the limiting value of the noise equivalent bandwidth as n approaches infinity?

(*Hint:* Note the definite integral

$$\int_0^\infty \frac{x^m}{x^n + a^n} \, dx = \frac{\pi a^{m-n+1}}{n \sin\left(\frac{m+1}{n}\pi\right)}, \quad \text{for } 0 < m+1 < n.)$$

- 5. Let X(t) be a wide-sense stationary process with zero mean, autocorrelation function  $R_X(\tau)$ , and power spectral density  $S_X(f)$ . We are required to find a linear filter with impulse response h(t), such that the filter output is X(t) when the input is white noise of zero mean and power spectral density  $N_0/2$ .
  - (a) Determine the condition that the impulse response h(t) must satisfy in order to achieve this requirement.
  - (b) What is the corresponding condition on the transfer function H(f) of the filter?
- 6. Consider a narrow-band noise n(t) with autocorrelation function  $R_N(\tau)$  and power spectral density  $S_N(f)$ . We have

$$n(t) = n_I(t)\cos(2\pi f_c t) - n_Q(t)\sin(2\pi f_c t).$$

(a) Give a full derivation that the power spectral densities of the in-phase and quadrature components are given by

$$S_{N_I}(f) = S_{N_Q}(f) = \begin{cases} S_N(f - f_c) + S_N(f + f_c), & -B \le f \le B \\ 0, & \text{elsewhere} \end{cases}$$

where it is assumed that  $S_N(f)$  occupies the frequency interval  $f_c - B \leq |f| \leq f_c + B$ , and  $f_c > B$ .

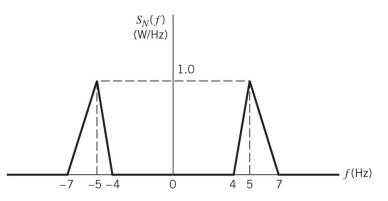
(b) Give a full derivation that the cross-spectral densities of the in-phase and quadrature components are given by

$$S_{N_I N_Q}(f) = -S_{N_Q N_I}(f) = \begin{cases} j \left[ S_N(f+f_c) - S_N(f-f_c) \right], & -B \le f \le B \\ 0, & \text{elsewhere.} \end{cases}$$

(*Hint:* You may use the results that

$$R_{N_{I}}(\tau) = R_{N_{Q}}(\tau) = R_{N}(\tau)\cos(2\pi f_{c}\tau) + \hat{R}_{N}(\tau)\sin(2\pi f_{c}\tau)$$
$$R_{N_{I}N_{Q}}(\tau) = -R_{N_{Q}N_{I}}(\tau) = R_{N}(\tau)\sin(2\pi f_{c}\tau) - \hat{R}_{N}(\tau)\cos(2\pi f_{c}\tau).$$

7. The power spectral density of a narrow-band noise n(t) is shown below. The carrier frequency is 5 Hz.



- (a) Plot the power spectral densities of the in-phase and quadrature components of n(t).
- (b) Plot their cross-spectral densities.
- 8. Consider a Gaussian noise n(t) with zero mean and power spectral density  $S_N(f)$  given by

$$S_N(f) = \begin{cases} N_0/2, & f_c - B \le f \le f_c + B\\ N_0/2, & -f_c - B \le f \le -f_c + B\\ 0, & \text{elsewhere.} \end{cases}$$

- (a) Let r(t) be the envelope of n(t) and R be the random variable obtained by observing r(t) at time  $t_1$ . Find the probability density function of R.
- (b) Find the mean and variance of R.

Homework Collaboration Policy: I allow and encourage discussion or collaboration on the homework. However, you are expected to write up your own solution and understand what you turn in. Late homework is subject to a penalty of 5% to 40% of your total points.