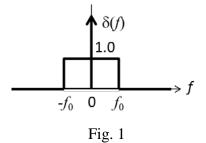
## Homework #7

1. Assume a noise  $n_1(t)$  is stationary with a power spectral density shown in Fig. 1. Another noise process  $n_2(t)$  is related with  $n_1(t)$  by:

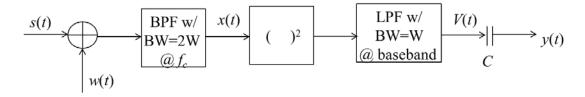
 $n_2(t) = n_1(t)\cos(2\pi f_c t + \theta) - n_1(t)\sin(2\pi f_c t + \theta),$ 

where  $f_c$  is a carrier frequency and  $\theta$  is the value of a random variable  $\Theta$  uniformly distributed within  $(0, 2\pi)$ .

- (a) Please find and sketch the autocorrelation function of  $n_1(t)$ .
- (b) Show that the cross correlation of  $n_1(t)$  and  $n_2(t)$  is 0. Assume that random variables  $N_1$  and  $\Theta$  are statistically independent.
- (c) Demonstrate that  $n_2(t) = \sqrt{2}n_1(t)\cos(2\pi f_c t + \pi/4 + \theta)$ .
- (d) Using (c) to find the autocorrelation function of  $n_2(t)$  in terms of the autocorrelation of  $n_1(t)$ :  $R_{N_1}(\tau)$ , where  $\tau$  is the time difference.
- (e) Plot power spectral density of  $n_2(t)$ .



- 2. Consider a phase modulation (PM) signal with the modulated wave defined by  $s(t) = A_c \cos[2\pi f_c t + k_p m(t)]$ , where  $k_p$  is a phase sensitivity and m(t) is the message. The filtered additive white Gaussian noise n(t) at the input of the phase detector is  $n(t) = n_I(t)\cos(2\pi f_c t) n_Q(t)\sin(2\pi f_c t)$ . Assume the carrier to noise ratio is large, please determine
  - (a) Output SNR
  - (b) FOM.
  - (c) Please compare this PM signal with FM one in terms of sinusoidal modulation.
- 3. A conventional AM signal is expressed as:  $s(t) = A_c [1 + \mu m(t)] \cos(2\pi f_c t)$ . Assume we apply a "square-law" detector to detect this signal with the following receiving block diagram:



- If  $m(t) = \cos(2\pi f_m t)$ .
- (a) Find  $SNR_C$ .
- (b) Find *SNR*<sub>*I*</sub>.
- (c) Please find V(t).
- (d) Please find y(t).

(e) Assume  $|\mu m(t)| \ll 1$ , please find  $SNR_O$ .

- (f) If  $SNR_C >> 1$ , Please find an approximation value of FOM.
- (g) If  $SNR_C \ll 1$ , Please find an approximation value of FOM. You may need the following Gaussian Integrals:

$$\int_{0}^{\infty} x^{2n} e^{-\frac{x^{2}}{a^{2}}} dx = \sqrt{\pi} \frac{a^{2n+1} (1 \times 3 \times 5 \dots \times (2n-1))}{2^{n+1}}$$
$$\int_{0}^{\infty} x^{2n+1} e^{-\frac{x^{2}}{a^{2}}} dx = \frac{n!}{2} a^{2n+2}$$

Please note: Homework must be turned in by the beginning of class. No late homework submission is allowable!

RN.IZ)  $S_{N_{i}}(f) = S(f) + \{1 | f| \le f.$ |.(a)|0 elsewhere  $\frac{R_{N_{i}}(t) = \overline{f_{i}}^{+} \left\{ S_{N_{i}}(f) \right\}}{= \int_{w}^{w} S_{N_{i}}(f) e^{j2\pi f t} df} \frac{1}{f} \\ = \int_{w}^{w} S(f) e^{j2\pi f t} df + \int_{f_{0}}^{f_{0}} 1 e^{j2\pi f t} df$ 韦 =  $1 + 2f_0 \operatorname{Sinc}(2f_0 T)$ *(b)*  $\frac{R_{N_1N_2}(\tau) = E[N_1(t+\tau) N_2(t)]}{E[N_1(t+\tau) N_2(t)]}$  $= \underline{F} \left[ N_1(t_1\tau) \cdot (N_1(t_1)cop(2\pi t_1\tau t_1 + \Theta) - N_1(t_1)sin(2\pi t_1\tau t_1 + \Theta) \right]$ = E[ NINHT) NILTI]. E[ Coo(27.fet+10)] -E[N, Ut+2) N, (t)]·E[Sin [=2, fet+0]]  $= RN_1(\tau) \cdot 0 - RN_1(\tau) \cdot 0 = 0$ N=1+)= n1+)co-12xfet+0)-n1+)sin (>xfet+0). (())=  $n_1(t) [w(2\pi f_c t + \theta) - sin(2\pi f_c t + \theta)]$ = vin(t)[ 症 (m(2xfct+b) - 症 sin(2xfct+b)] = VZNIH/[Coo 柔 Coo 127 fet++) - sin 歪 sinl=x fet++) = J2 n.(t) wo (2xtit+ &+ 0). (d)  $R_{N_2}(\tau) = E[N_2(t+\tau), N_2(t)]$ = 2 E[N(1+12).N(1+1] E[ (~(27 f(1+12)+2+0) (~)27 f(1+2+0)] => RN, (2) + (10 (27, fct) = RN, (2) - Cool 27, fct). SN2(+) = FSRN2(2)3 = FSRN1(2) C= 122 fc2)3 (e)  $= \pm SN_1 (f - f_L) + \pm SN_1 (f + f_L)$ 38(f+f,) SN2(+) -28(f-f.) -te-to-fe-tetto o -fe-to-fe-tetto f

For a PM signal: sit) = Ac coo(2xfit + \$t)), where \$t) = kp mit) 2.  $\gamma(t) = s(t) + n(t) = Ac \cos(2\pi f_c t + g(t)) + \gamma(t) \cos(2\pi f_c t + y(t))$ ; where  $r(t) = [n_1^2(t) + n_2^2(t)]^{1/2}$ ,  $\psi(t) = tan^{-1}(\frac{n_0(t)}{n_1(t)})$ the received phase  $\theta(t) = \varphi(t) + t_{qn} - \frac{1}{2} \frac{r(t) \sin(\psi(t) - \varphi(t))}{Ac + r(t) \cos(\psi(t) - \varphi(t))}$ For large CNR,  $A_{c} >> r(t)$   $O(t) \simeq \phi(t) + \frac{r(t)}{\delta} \sin(\psi(t) - \phi(t))$  $= kpm(t) + \frac{\gamma(t)}{4}sin(\psi(t) - \beta(t))$ Since Ult) is uniformly distributed over [0,27], so is 1/1)- \$1, => The noise after phase detector is independent of  $\phi(t) \Rightarrow$  The output  $\gamma(t) = k_p m(t) + \gamma(t) \sin(\gamma(t)) = k_p m(t) + \frac{n_0 t}{A_c}$ The output signal is  $k_p m(t) \Rightarrow$  signal power  $P_s = k_p^2 P$ . Noise power:  $P_N = \frac{1}{A_c^2} \cdot N_0 \cdot 2W \Rightarrow (SNR)_0 = A_c^2 k_p^2 P_2 N_0 W$ (a)The channel SNR: the modulated signal power: Ps = 1/2 Ac<sup>2</sup> The noise power = N° × 2W = NoW. .: (SNP)c = Ac<sup>2</sup>/2NoW (6) : FOM= (SNR)0/(SNR)c = kp2P (C)It the message is a sinuspidal wave : mut) = Am Coolarfint) P= 1 Am, Then (FOM) pM = = (kp/1m) = = Bp, where Bp is phase deviation in PM. (FOM)==== B<sup>2</sup>, where B is the phase deviation. in FM. . For the same phase deviation, FM is 3 times better than PM. S(t) = ACCI+MMUT) ] Cool2xfit) and MUTI= cool2xfmt) 3  $\frac{\text{then signal power: } P_{s} = \frac{1}{2} Ac^{2} (1 + M^{2} z)}{(01)} (SNR)_{c} = \frac{1}{2} Ac^{2} (1 + M^{2} z)}{N_{0} W} = \frac{Ac^{2} (2 + M^{2})}{4N_{0} W}$  $\frac{(b)}{(SNP)_{I}} = \frac{\frac{1}{2}Ac^{2}(1+M^{2}/2)}{2NOW} = \frac{Ac^{2}(2+M^{2})}{8NOW}$ 

 $(c) |_{V(t)} = LPF \{q^{2}(t)\} = LPF \{ (s(t) + n(t))^{2} \}, if u[m(t)] < 1$  $\simeq A^{c^2}[1+2 Um(t)] + A^{c}[1+Um(t)] n_{t}(t) + \frac{1}{2}n_{t}^2(t) + \frac{1}{2}n_{a}(t)$ ytt) is the time varying term of V(t) ⇒ y(t) = Ac<sup>2</sup> Um(t) + Ac [ It Um(t)] nI(t) + ± NJ<sup>2</sup>(t) + ± NB(t) d) | demodulated message power: PM = + Act M2 demodulated noise: no= Ac[1+Mmuti] n=1(t) + = n=2(t) 191 The corresponding noise power: PN = Ono = E[no] - (E[no]) = PN = Ac (H = ) GN + GN where ON = = NoW  $= \frac{1}{(S_{1}^{2}N_{0}^{2})} = \frac{1}{2} \frac{Ac^{4}M^{2}}{Ac^{2}(1+\frac{M^{2}}{2})} = \frac{2Ac^{4}M^{2}}{(Ac^{2}(2+M^{2})+26N^{2})} = \frac{2Ac^{4}M^{2}}{(Ac^{2}(2+M^{2})+26N^{2})} = \frac{1}{26N^{2}}$  $= 2\left(\frac{M}{2+M^2}\right)^2 \left(\frac{(SNR)_c}{1+\frac{1}{(SNR)_c}}\right) - where \left(\frac{SNR}{c}\right) = \frac{Ac^2(2+M^2)}{4N_0W}$  $\frac{if(SNR)c77}{if(SNR)c<<|(SNR)o~2(\frac{M}{S+M^2})(SNR)c}$ (F) (91  $n(t) = n_1(t) \cos(2\pi f_c t) - n_0(t) \sin(2\pi f_c t) = r(t) \cos(2\pi f_c t + \psi(t))$ Note: where vit)= Nit)+Natt), Ut)= tan-1 north  $Var[\pm N_{1}^{2} + \pm N_{0}^{2}] = Var[\pm Y^{2}] = E[(\pm r^{2})^{2}] - (E[\pm r^{2}])^{2} = \pm [E[r^{4}] - (E[r^{2}])^{2}]$  $\frac{VHL2THS}{The Pdf of R is f_{R}(Y) = \frac{Y}{6\pi}e^{-\frac{Y^{2}}{26\pi}} and o \leq Y \leq \infty}{Fyom Gaussian Integral: <math display="block">\int_{0}^{\infty} \chi^{2n+1}e^{-\frac{T^{2}}{4\pi}} dX = \frac{n!}{2}a^{2n+2}$   $\frac{E[Y^{4}] = \int_{0}^{\infty} Y^{4} \cdot \frac{Y}{6\pi}e^{-\frac{Y^{2}}{26\pi}} dY = \frac{f_{2}}{5\pi}\int_{0}^{\infty} Y^{5}e^{-\frac{Y^{2}}{26\pi}} dY, take h=2, a=\sqrt{2}6M$   $= \frac{1}{6\pi} \cdot \frac{Z'}{5\pi} (\sqrt{2}6M)^{6} = 86M$   $E[Y^{2}] = \int_{0}^{\infty} Y^{2} \cdot \frac{Y}{6\pi}e^{-\frac{Y^{2}}{26\pi}} dY = \frac{1}{6\pi^{2}} (\sqrt{2}6M)^{7} = 26M^{2}$ :  $Var[\pm n_{\tilde{t}}^{2}+\pm n_{\tilde{t}}^{2}]=\pm [E[r^{4}]-(E[r^{2}])^{2}]=\pm [86N^{4}-126N^{2}]^{2}$ = = + + + 5N = 5N