

1.

(a) $S(t) = A_c \cos(2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau)$

(b) $S(t) = \operatorname{Re} \{ S(t) e^{j2\pi f_c t} \} = \operatorname{Re} \{ A_c e^{j2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau} \}$
 $= \operatorname{Re} \{ A_c e^{j2\pi k_f \int_0^t m(\tau) d\tau} e^{j2\pi f_c t} \}$

$\therefore S(t) = A_c e^{j2\pi k_f \int_0^t m(\tau) d\tau}$.

(c) Within weak modulation scheme, we can let $|2\pi k_f \int_0^t m(\tau) d\tau| \ll 1$.

$$\therefore S(t) = A_c e^{j2\pi k_f \int_0^t m(\tau) d\tau} \approx A_c [1 + j2\pi k_f \int_0^t m(\tau) d\tau]$$

$$\therefore S(t) = \operatorname{Re} \{ S(t) e^{j2\pi f_c t} \} \approx \operatorname{Re} \{ A_c (1 + j2\pi k_f \int_0^t m(\tau) d\tau) e^{j2\pi f_c t} \}$$

$$= A_c \cos(2\pi f_c t) - A_c \cdot 2\pi k_f \int_0^t m(\tau) d\tau \cdot \sin(2\pi f_c t)$$

(d) Since the information is $m(t)$ is carried by $\sin(2\pi f_c t)$, we have to apply $c(t) = -2 \sin(2\pi f_c t)$ to the product modulator.

Then $r(t) = s(t) \cdot c(t)$.

$$= [A_c \cos(2\pi f_c t) - A_c \cdot 2\pi k_f \int_0^t m(\tau) d\tau \cdot \sin(2\pi f_c t)] \times (-2 \sin(2\pi f_c t))$$

$$= -2 A_c \sin(2\pi f_c t) \cos(2\pi f_c t) + 2 A_c \cdot 2\pi k_f \int_0^t m(\tau) d\tau \cdot \sin^2(2\pi f_c t)$$

$$= -A_c \cdot \sin(4\pi f_c t) + A_c \cdot 2\pi k_f \int_0^t m(\tau) d\tau$$

$$- A_c \cdot 2\pi k_f \int_0^t m(\tau) d\tau \cdot \cos(4\pi f_c t)$$

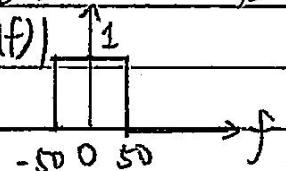
After the LPF, we have $A_c \cdot 2\pi k_f \int_0^t m(\tau) d\tau = 75$

After the differentiator, we have $A_c \cdot 2\pi k_f \cdot m(t)$

(e) $m(t) = \frac{\sin(100\pi t)}{\pi t} = 100 \operatorname{sinc}(100t)$ bandwidth = 50 Hz

Take Fourier Transform on $m(t)$ we get $M(f) = \operatorname{rect}\left(\frac{f}{100}\right)$

(f) From the Rayleigh's Energy theorem $\int_{-\infty}^{\infty} |m(t)|^2 dt = \int_{-\infty}^{\infty} |M(f)|^2 df$

$$\because M(f) = \operatorname{rect}\left(\frac{f}{100}\right) \Rightarrow |M(f)|^2$$


$$\therefore E = \int_{-\infty}^{\infty} |m(t)|^2 dt$$

$$= \int_{-\infty}^{\infty} |M(f)|^2 df$$

$$= \int_{-50}^{50} |\operatorname{rect}\left(\frac{f}{100}\right)|^2 df = 100 \text{ joule.}$$

(g) $s(t) = 3 \cos(2\pi 10^6 t + 100 \pi \int_0^t m(\tau) d\tau)$
 $= A_c \cos(2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau) \Rightarrow 2\pi k_f = 100\pi \Rightarrow k_f = 50$
 $m(t) = 100 \sin(100\pi t) \Rightarrow \max\{m(t)\} = 100$
 $\therefore sf = k_f \cdot \max\{m(t)\} = 50 \times 100 = 5000$,
maximum phase deviation $\beta = \frac{\Delta f}{W} = \frac{5000}{50} = 100 \gg 1$.
It's not a weak FM signal.

(h) Carson's rule: $B_T = 2(\Delta f + W) = 2(5000 + 50) = 10.1 \text{ kHz}$

2.

(a) $o_i(t) = 2\pi f_c t + k_p m(t) = 2\pi f_c t + k_p A_m \cos(2\pi f_m t)$
 $= 2\pi f_c t + \beta_p \cos(2\pi f_m t)$ where β_p is phase deviation.

Then the instantaneous frequency is

$$f_i(t) = \frac{d o_i(t)}{dt} = f_c - \beta_p f_m \sin(2\pi f_m t)$$

\therefore the frequency deviation $|\Delta f| = \beta_p f_m$

$$\text{By Carson's rule: } B_T = 2\Delta f \left(1 + \frac{1}{\beta_p}\right)$$

$$\because \beta_p \gg 1 \Rightarrow \frac{1}{\beta_p} \ll 1 \Rightarrow B_T \approx 2\Delta f = 2\beta_p f_m \propto f_m$$

\therefore The bandwidth is linearly proportional to the modulation frequency f_m

(b) In FM, $\Delta f = k_f A_m$, $B_T = 2\Delta f + 2f_m$. Since Typically Δf is much larger than f_m , the transmission bandwidth is mainly dominated by the amplitude A_m or frequency sensitivity k_f and is not related to modulation frequency.

3.

$$s(t) = 10 \cos(2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau)$$

$$f_c = 100 \text{ MHz}, m(t) = 10 \cos(2\pi f_m t), f_m = 3 \text{ kHz}$$

(a) After the differentiator:

$$s'(t) = -10(2\pi f_c + 2\pi k_f m(t)) \sin(2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau)$$

For the envelope detector, $2\pi f_c + 2\pi k_f m(t)$ should be always positive $\Rightarrow f_c + k_f 10 \cos(2\pi f_m t) \geq 0 \Rightarrow f_c - 10k_f \geq 0$

$$\Rightarrow k_f \leq f_c/10 = 100 \times 10^6 / 10 = 10^7 \text{ Hz/Volt}$$

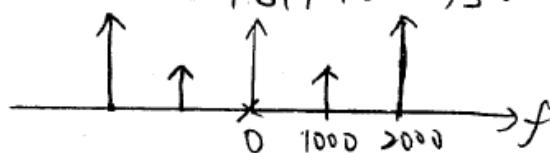
- (b) $k_f = 10$ From Carson's rule $B_T = 2\Delta f + 2f_m$
 $\Delta f = k_f \cdot A_m = 10 \cdot 10 = 100$ and $f_m = 3 \text{ kHz} \Rightarrow B_T = 6.2 \text{ kHz}$
 $\beta = \Delta f / f_m = \frac{100}{3000} \ll 1 \Rightarrow$ This is NBFM.
- (c) The instantaneous frequency is $f_c + k_f m(t) = 10^8 + 100 \cos(2\pi f_m t)$
∴ The maximum frequency is $10^8 + 100 \text{ Hz}$.
The minimum frequency is $10^8 - 100 \text{ Hz}$
- (d) For an envelope detector to work properly, the charging time constant $R_s C$ must be short compared to the carrier period $1/f_c$. Thus $R_s \ll 1/f_c = 1/10^8 \cdot 10^9 = 10 \Omega$. choose $R_s = 0.1 \Omega$. On the other hand, the discharging time constant should be much greater than the carrier period $1/f_c = 10^{-8} \text{ sec}$ but much smaller than the message period of $1/3000 \text{ sec}$. So we may pick R_d to be 1000Ω , so that the discharging time constant is $R_d \cdot C = 10^3 \cdot 10^{-9} = 10^{-6} \text{ sec}$.
- (e) To find the minimum interval T such that there are at least 4 zero crossings, we must consider the minimum value of the instantaneous frequency $f(t)$. The minimum value is $10^8 - 100$. Since there are two zero crossing in every carrier period,
 $\frac{4}{2T} = 10^8 - 100 \Rightarrow T = \frac{2}{10^8 - 100} = 2.00002 \times 10^{-8} \approx 2 \times 10^{-8} \text{ sec}$.

4.

(a) $s(t) = A_c \cos(2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau)$
 $= 10 \cos(2\pi f_0 t + 10 \sin(2\pi \cdot 1000 t) + 20 \sin(2\pi \cdot 2000 t))$
 $\therefore 2\pi k_f \int_0^t m(\tau) d\tau = 10 \sin(2\pi \cdot 1000 t) + 20 \sin(2\pi \cdot 2000 t)$

Take the derivative on both sides

 $2\pi k_f m(t) = 2\pi \cdot 10^4 \cos(2\pi \cdot 1000 t) + 2\pi \cdot 4 \times 10^5 \cos(2\pi \cdot 2000 t)$
 $\Rightarrow m(t) = \frac{10^4}{k_f} (\cos(2\pi \cdot 1000 t) + 4 \cos(2\pi \cdot 2000 t))$
 $\Rightarrow M(f) = \frac{1}{2} \frac{10^4}{k_f} [\delta(f - 1000) + \delta(f + 1000) + 4 \delta(f - 2000) + 4 \delta(f + 2000)]$



$\therefore W = 2000 \text{ Hz} = 2 \text{ kHz}$

$$(b) \parallel P_{av} = \frac{1}{2} A c^2 = \frac{1}{2} \cdot 10^2 = 50 \text{ W}$$

(c)

The instantaneous frequency is

$$f_i(t) = \frac{1}{2\pi} \frac{d}{dt} [2\pi f_0 t + 10 \sin(2\pi \cdot 1000t) + 20 \sin(2\pi \cdot 2000t)] \\ = 10^6 + 10^4 \cos(2\pi \cdot 1000t) + 4 \times 10^4 \cos(2\pi \cdot 2000t).$$

(d)

When $\cos(2\pi \cdot 1000t) = \cos(2\pi \cdot 2000t) = 1$, we have maximum frequency deviation $10^4 + 4 \times 10^4 = 50 \text{ kHz}$.

and when $2\pi \cdot 1000t_n = k \cdot 2\pi, 2\pi \cdot 2000 = l \cdot 2\pi, k, l: \text{integer}$

$$t_n = \frac{k}{1000} = \frac{l}{2000} \Rightarrow l = 2k$$

\Rightarrow when $t_n = k \cdot 10^{-3} \text{ sec } k = \text{integer}$, we have maximum sf.

(e)

$$B_T = 2(\Delta f + W) = 2(50k + 2k) = 104 \text{ kHz.}$$